

Section 25

Bipolar Junction Transistor – Design

25.5 Poly-Si emitter

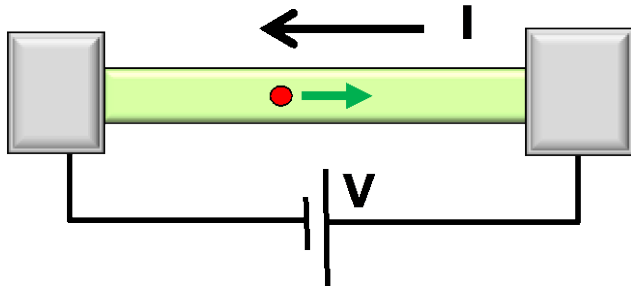
Gerhard Klimeck
gekco@purdue.edu



School of Electrical and
Computer Engineering

Section 25

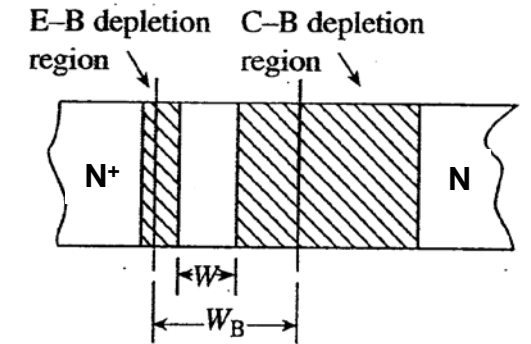
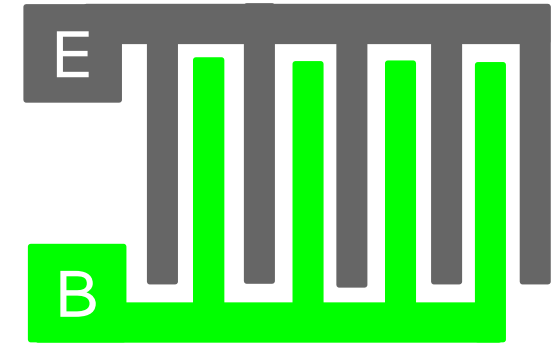
Bipolar Junction Transistor - Design



$$I = G \times V$$

$$= q \times n \times v \times A$$

↑ charge density
 ↑ velocity
 area



> • 25.1 Current gain in BJTs

> • 25.2 Base Doping Design

- » Current Crowding – Non-Uniform Turn-On
- » Punch-through
- » Base Width Modulation

> • 25.3 Collector Doping Design (Kirk Effect, Base Pushout)

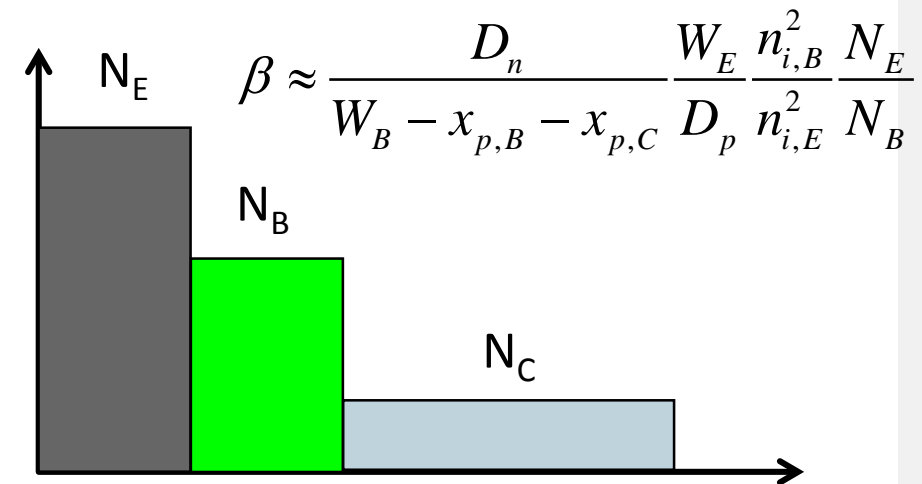
> • 25.4 Emitter Doping Design

> • 25.5 Poly-Si emitter

> • 25.6

$$x_{p, BE} = \sqrt{\frac{2k_s \epsilon_0}{q} \frac{N_E}{N_B (N_E + N_B)} (V_{bi} - V_{BE})}$$

$$x_{p, BC} = \sqrt{\frac{2k_s \epsilon_0}{q} \frac{N_C}{N_B (N_C + N_B)} (V_{bi} - V_{BC})}$$

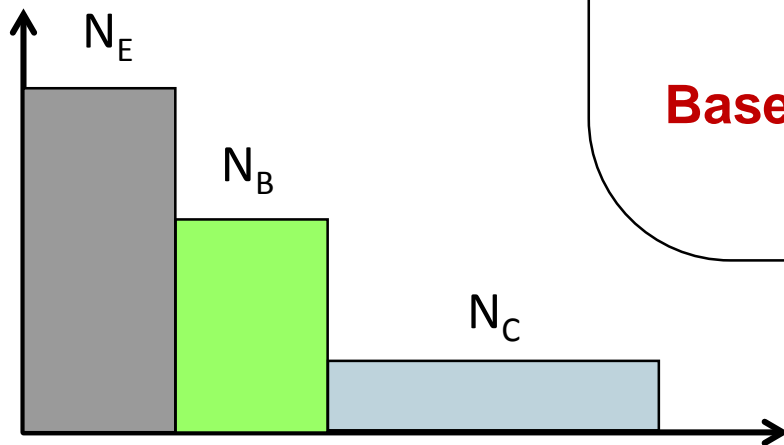


$$V_A = -\frac{q N_B W_B}{C_{CB}} \quad C_{CB} = \frac{K_s \epsilon_0}{x_{n,C} + x_{p,B}}$$

Doping for Gain

$$\beta_{DC} = \frac{I_C}{I_B}$$

$$\beta_{dc} \approx \frac{D_n W_E n_{i,B}^2 N_E}{W_B D_p n_{i,E}^2 N_B}$$

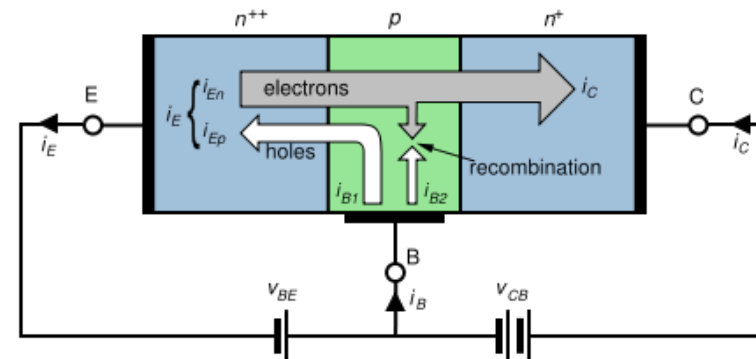


Emitter doping: As high as possible without *band gap narrowing*

Base doping: As low as possible, without *current crowding, Early effect*

Collector doping: Lower than base doping *without Kirk Effect*

Base Width: As thin as possible without *punch through*



How to make better Transistor

$$\beta_{DC} = \frac{I_C}{I_B}$$

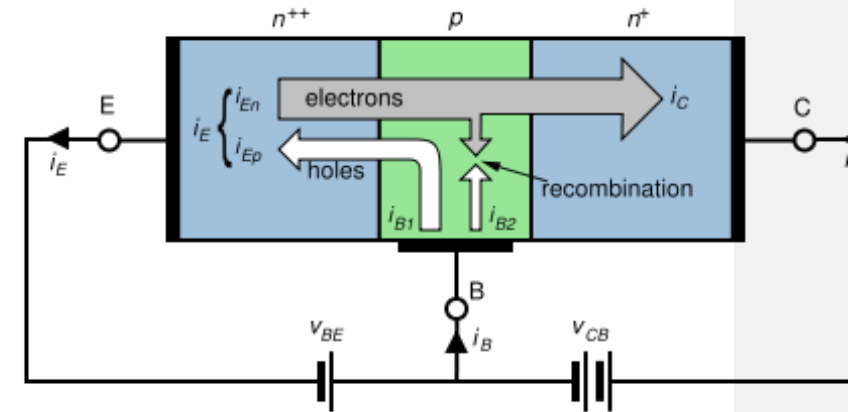
Polysilicon Emitter
 suppress D_p
 suppress base current

$$\beta_{dc} \approx \frac{D_n W_E n_{i,B}^2 N_E}{W_B D_p n_{i,E}^2 N_B}$$

Graded Base transport
 (build in electric field to accelerate el)

Classical Shockley Transistor

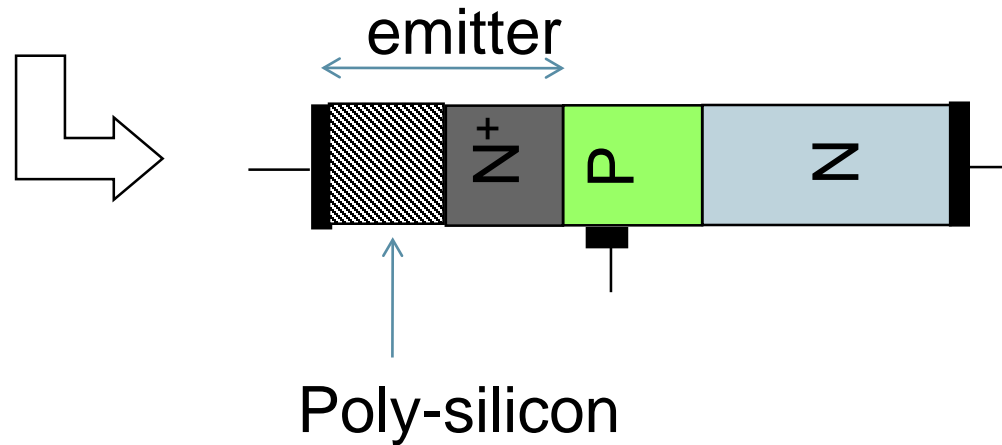
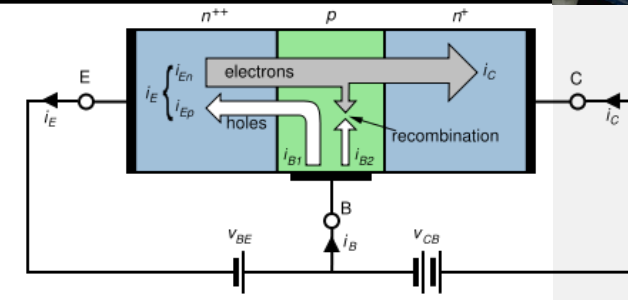
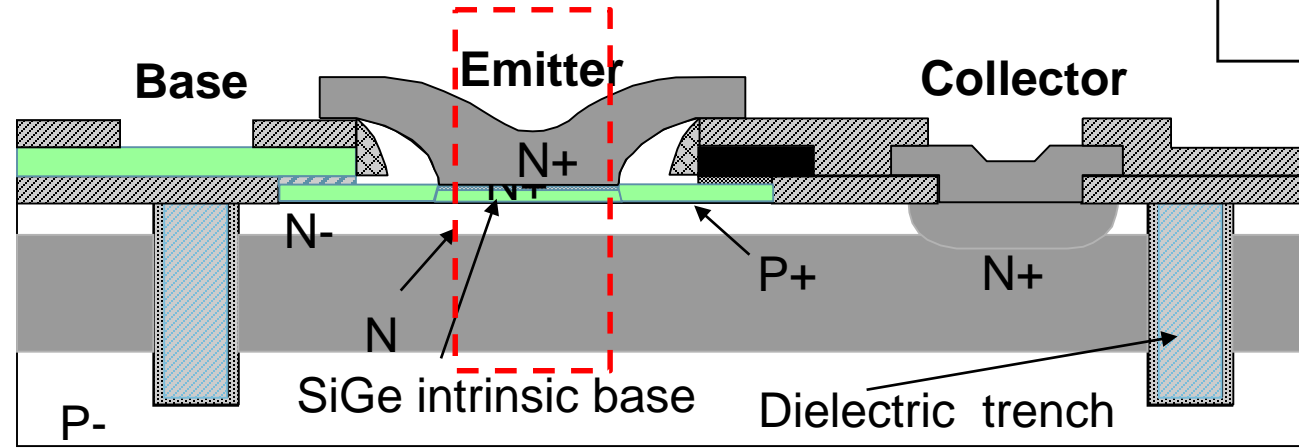
Hetero-junction Bipolar Transistor
 Nobel Prize Kroemer/Alferov



Poly-silicon Emitter

$$\beta_{DC} = \frac{I_C}{I_B}$$

Polysilicon Emitter
 suppress D_p
 suppress base current



Poly-silicon Emitter

$$\beta_{DC} = \frac{I_C}{I_B}$$

Polysilicon Emitter
suppress D_p
suppress base current

$$I_{p,E,si} = -qD_p \frac{dp}{dx}$$

$$I_{p,E,si} = -q \left(\frac{D_p}{W_E} \right) p_1$$

$$I_{p,E,poly} = -qD_p \frac{dp}{dx}$$

$$I_{p,E,poly} = -qD_p \frac{p_1 - p_2}{W_E}$$

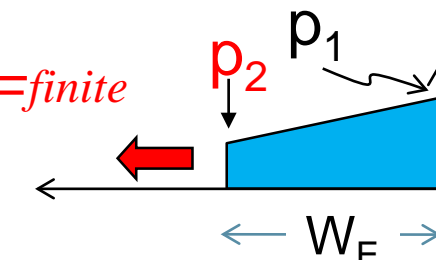
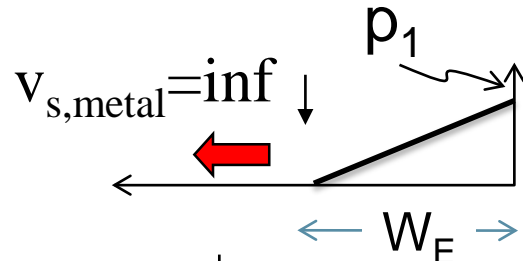
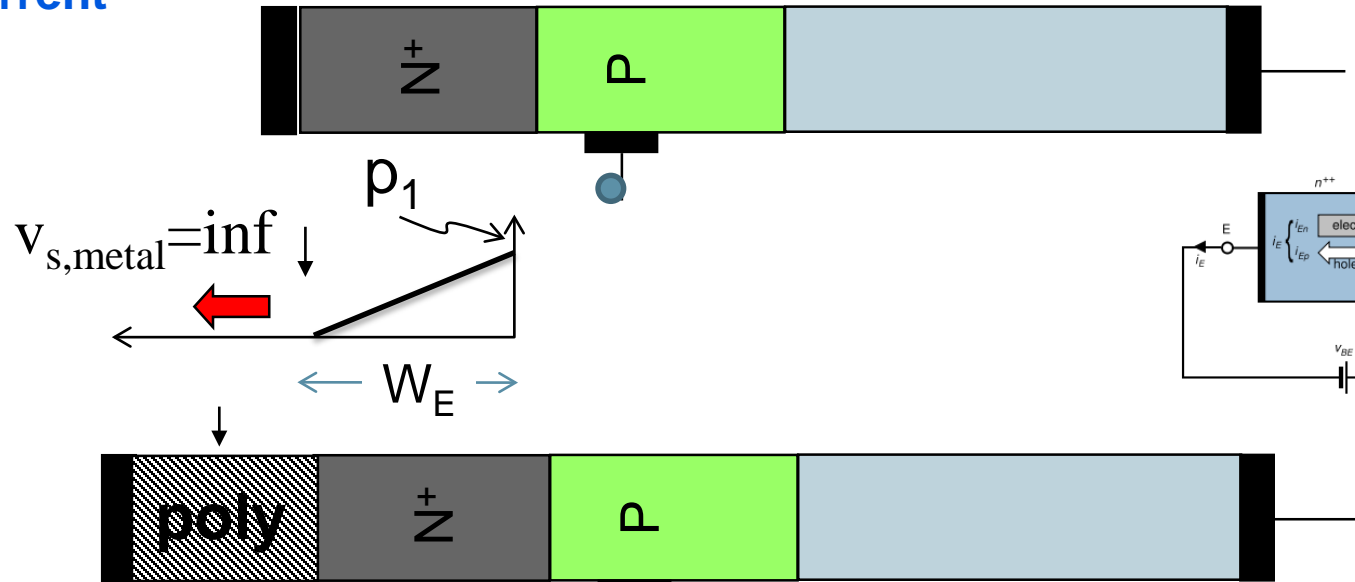
$$I_{p,E,poly} = -qu_s p_2$$

$$p_2 = p_1 \frac{\frac{D_p}{W_E}}{\frac{D_p}{W_E} + u_s}$$

$$\frac{I_{p,E,poly}}{I_{p,E,si}} = \frac{qu_s p_2}{q \left(\frac{D_p}{W_E} \right)}$$

$$\frac{I_{p,E,poly}}{I_{p,E,si}} = \frac{u_s}{\frac{D_p}{W_E} + u_s}$$

$$\frac{I_{p,E,poly}}{I_{p,E,si}} = \frac{1}{1 + \frac{D_p}{W_E} \frac{1}{u_s}}$$



$v_{s,poly} = \text{finite}$ reduces $I_{p,E,poly}$

Poly-silicon Emitter

$$\beta_{DC} = \frac{I_C}{I_B}$$

Polysilicon Emitter
suppress D_p
suppress base current

$$I_{p,E,si} = -qD_p \frac{dp}{dx}$$

$$I_{p,E,si} = -q \left(\frac{D_p}{W_E} \right) p_1$$

$$I_{p,E,poly} = -qD_p \frac{dp}{dx}$$

$$I_{p,E,poly} = -qD_p \frac{p_1 - p_2}{W_E}$$

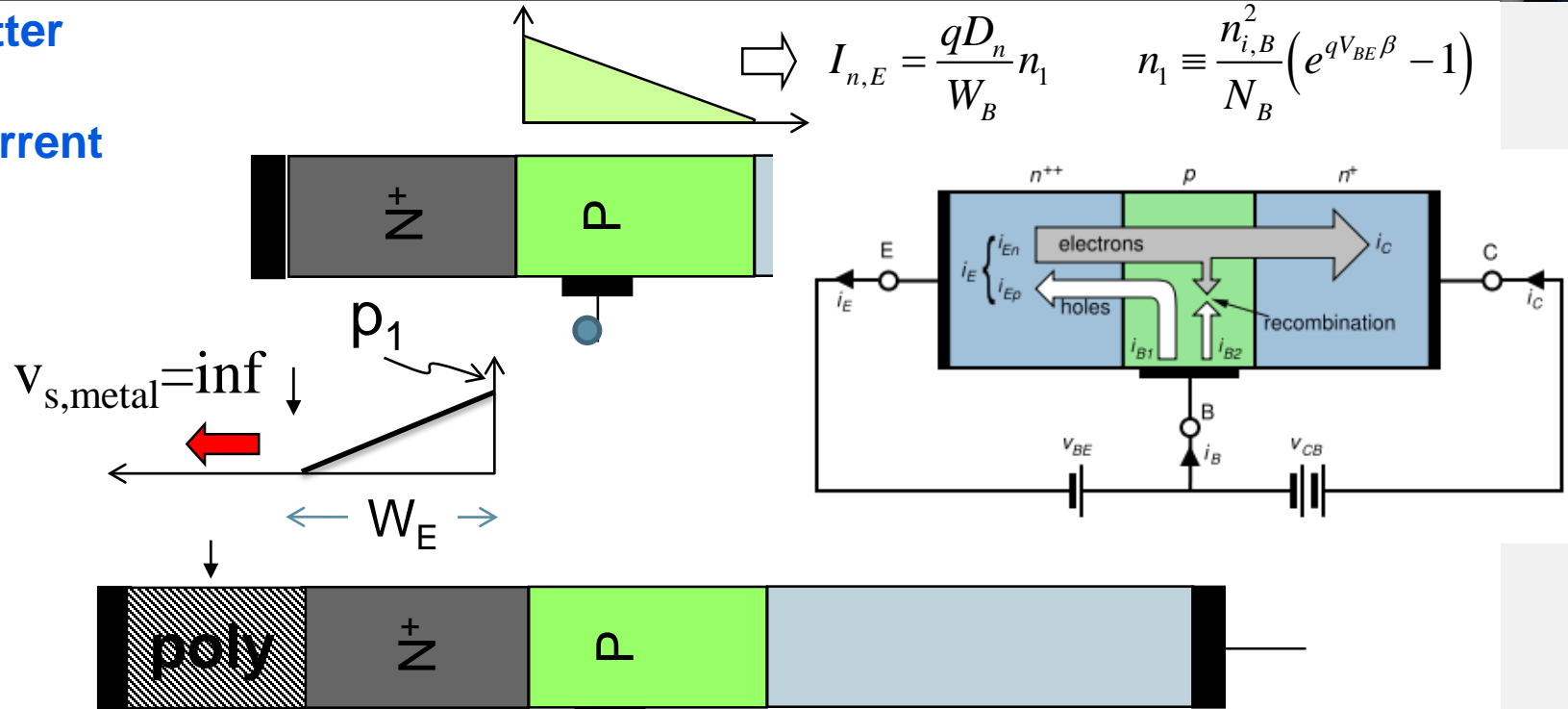
$$I_{p,E,poly} = -q v_s p_2$$

$$p_2 = p_1 \frac{\frac{D_p}{W_E}}{\frac{D_p}{W_E} + v_s}$$

$$\frac{I_{p,E,poly}}{I_{p,E,si}} = \frac{q v_s p_2}{q \left(\frac{D_p}{W_E} \right)}$$

$$\frac{I_{p,E,poly}}{I_{p,E,si}} = \frac{v_s}{\frac{D_p}{W_E} + v_s}$$

$$\frac{I_{p,E,poly}}{I_{p,E,si}} = \frac{1}{1 + \frac{D_p}{W_E} \frac{1}{v_s}}$$



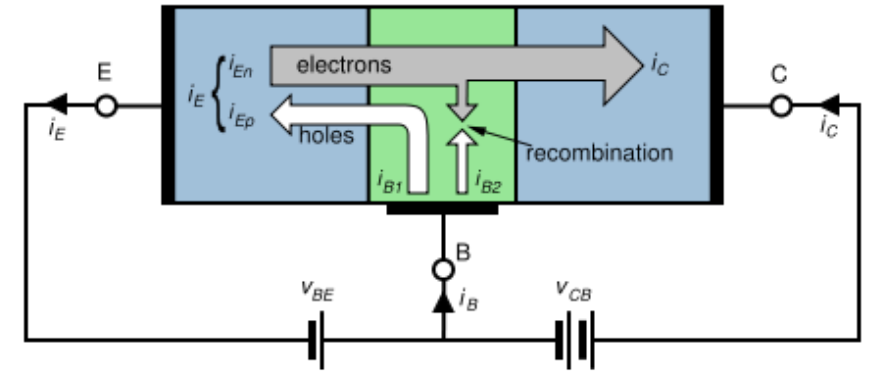
$$I_{n,E} = \frac{qD_n}{W_B} n_1 \quad n_1 \equiv \frac{n_{i,B}^2}{N_B} (e^{qV_{BE}\beta} - 1)$$

$v_{s,poly} = \text{finite}$

$v_{s,poly} = \text{finite}$ reduces $I_{p,E,poly}$

Gain in Poly-silicon Transistor

$$\frac{I_{p,E,poly}}{I_{p,E,Si}} = \frac{v_s}{\frac{D_p}{W_E} + v_s} \quad \frac{I_{B,poly}}{I_{B,Si}} = \frac{v_s}{\frac{D_p}{W_E} + v_s}$$



$$\beta_{poly} = \frac{I_C}{I_{B,poly}} = \left(\frac{I_C}{I_{B,Si}} \right) \times \left[\frac{I_{B,Si}}{I_{B,poly}} \right] \approx \left(\frac{D_n}{W_B} \frac{W_E}{D_p} \frac{n_{i,B}^2}{n_{i,E}^2} \frac{N_E}{N_B} \right) \times \left[\frac{D_p/W_E + v_s}{v_s} \right]$$

$$\beta_{poly} \sim \frac{D_n}{W_B} \frac{n_{i,B}^2}{n_{i,E}^2} \frac{N_E}{N_B} \times \frac{1}{v_s}$$

$$v_s \ll \frac{D_p}{W_E}$$

Poly suppresses base current, increases gain ...

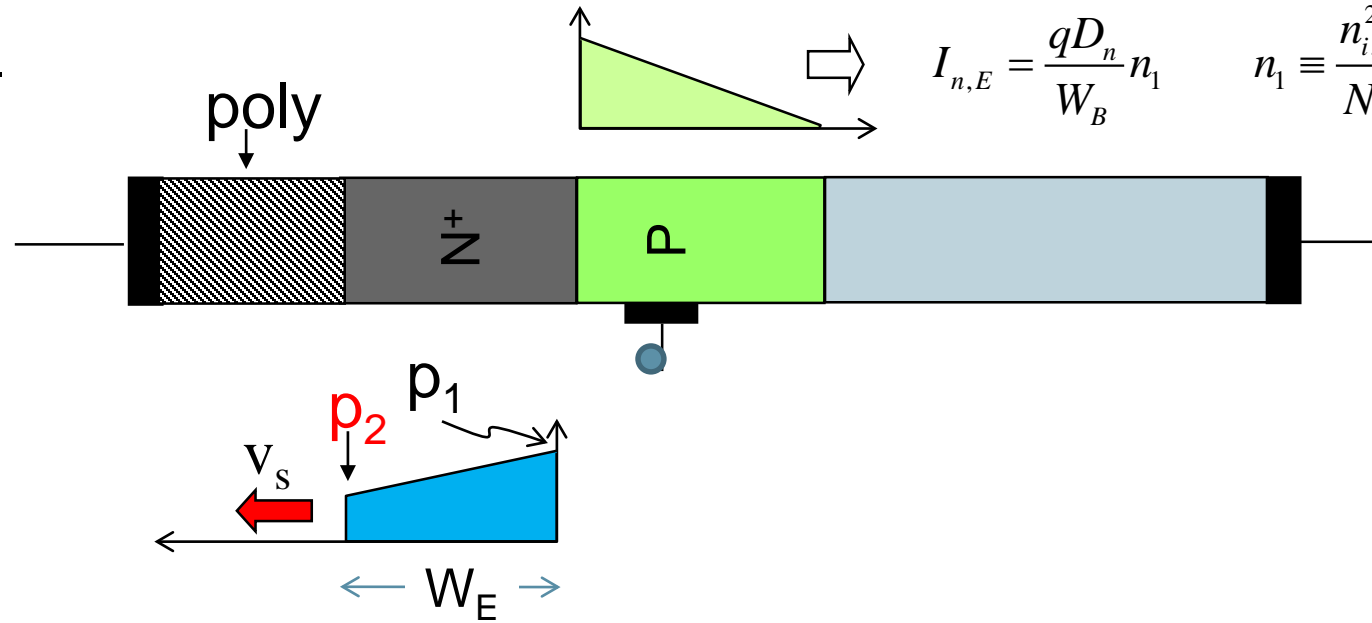
$$I_{n,E} = \frac{qD_n}{W_B} n_1 \quad n_1 \equiv \frac{n_{i,B}^2}{N_B} (e^{qV_{BE}\beta} - 1)$$

Question: Why does poly only suppress the hole current, not electron current?

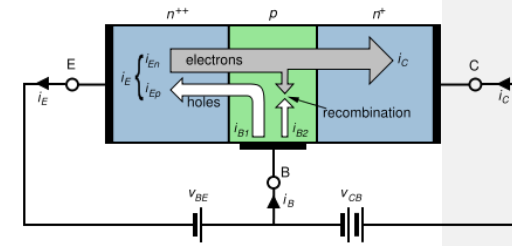
Ans. Polysilicon is not a ohmic contact and acts as rectifying contact. It blocks the easy passage of holes but lets electrons pass through

Poly-silicon Emitter

$$\frac{I_{B,poly}}{I_{B,Si}} = \frac{v_s}{\frac{D_p}{W_E} + v_s}$$



$$I_{n,E} = \frac{qD_n}{W_B} n_1 \quad n_1 \equiv \frac{n_{i,B}^2}{N_B} (e^{qV_{BE}\beta} - 1)$$

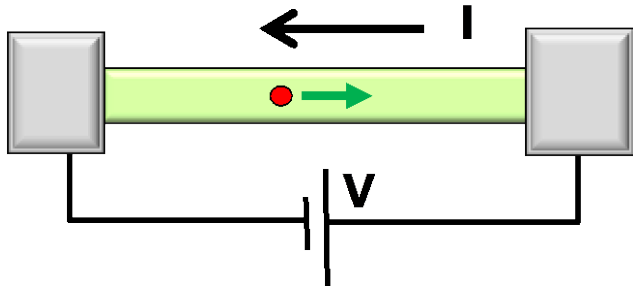


$$\beta_{poly} = \frac{I_C}{I_{B,poly}} = \left(\frac{I_C}{I_{B,Si}} \right) \times \left[\frac{I_{B,Si}}{I_{B,poly}} \right] \approx \left(\frac{D_n}{W_B} \frac{W_E}{D_p} \frac{n_{i,B}^2}{n_{i,E}^2} \frac{N_E}{N_B} \right) \times \left[\frac{D_p/W_E + v_s}{v_s} \right]$$

$$\beta_{poly} \sim \frac{D_n}{W_B} \frac{n_{i,B}^2}{n_{i,E}^2} \frac{N_E}{N_B} \times \frac{1}{v_s}$$

Section 25

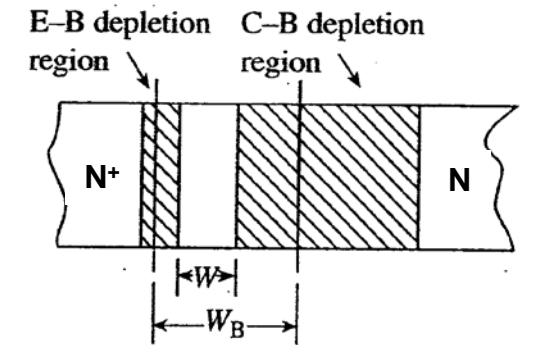
Bipolar Junction Transistor - Design



$$I = G \times V$$

$$= q \times n \times v \times A$$

↑ charge density ↑ velocity area



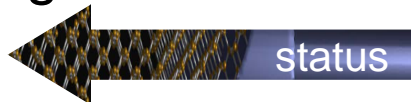
$$x_{p, BE} = \sqrt{\frac{2k_s \epsilon_0}{q} \frac{N_E}{N_B (N_E + N_B)} (V_{bi} - V_{BE})}$$

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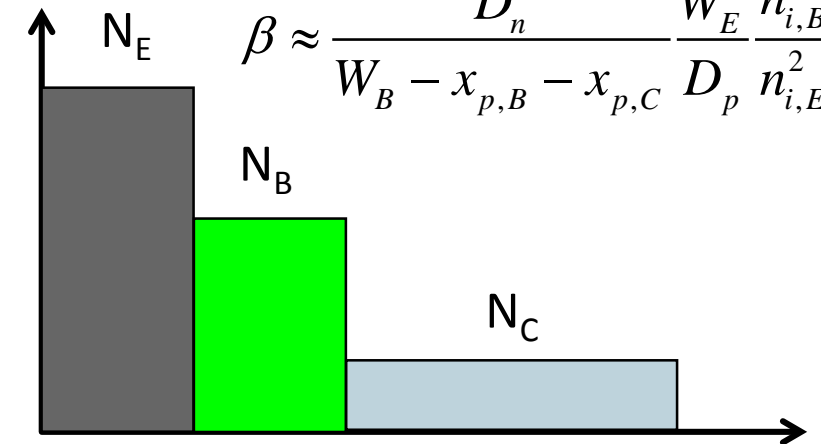
$$\beta \approx \frac{D_n}{W_B - x_{p,B} - x_{p,C}} \frac{W_E n_{i,B}^2 N_E}{D_p n_{i,E}^2 N_B}$$

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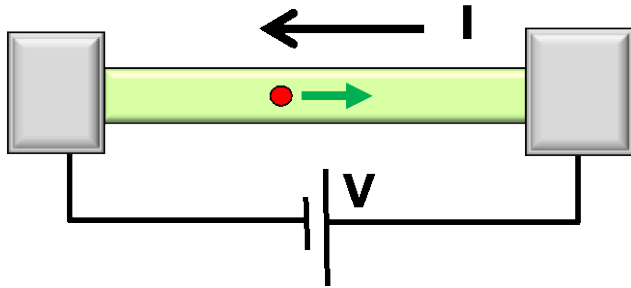


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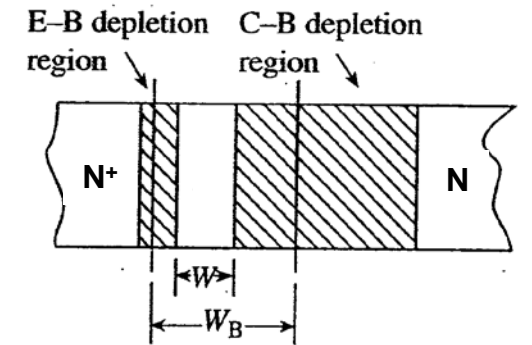
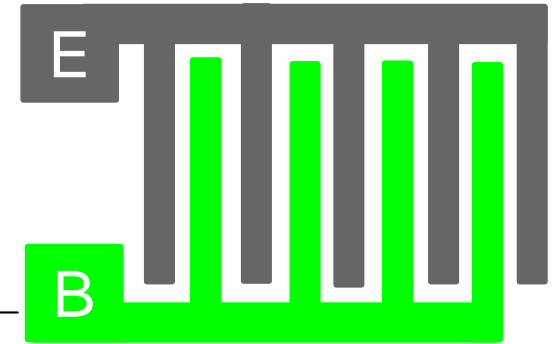
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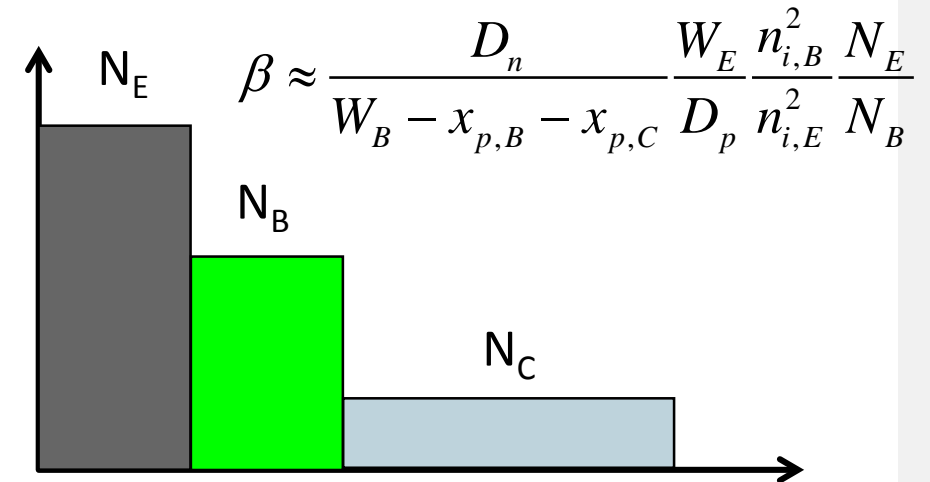
$$= q \times n \times v \times A$$

↑ charge density
 ↑ velocity
 area



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$$x_{p, BC} = \sqrt{\frac{2k_s \epsilon_0}{q} \frac{N_C}{N_B (N_C + N_B)} (V_{bi} - V_{BC})}$$



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- > • 25.4 Emitter Doping Design
- > • 25.5 Poly-Si emitter
- > • 25.6 Short base transport



$$V_A = -\frac{qN_B W_B}{C_{CB}}$$

$$C_{CB} = \frac{K_s \epsilon_0}{x_{n,C} + x_{p,B}}$$