

Section 24

Bipolar Junction Transistor - Fundamentals

24.3 Currents in BJTs

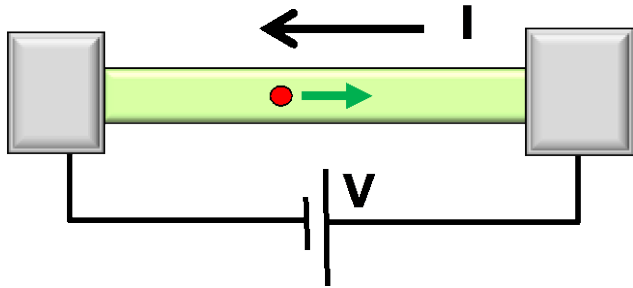
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School of Electrical and
Computer Engineering

Section 24

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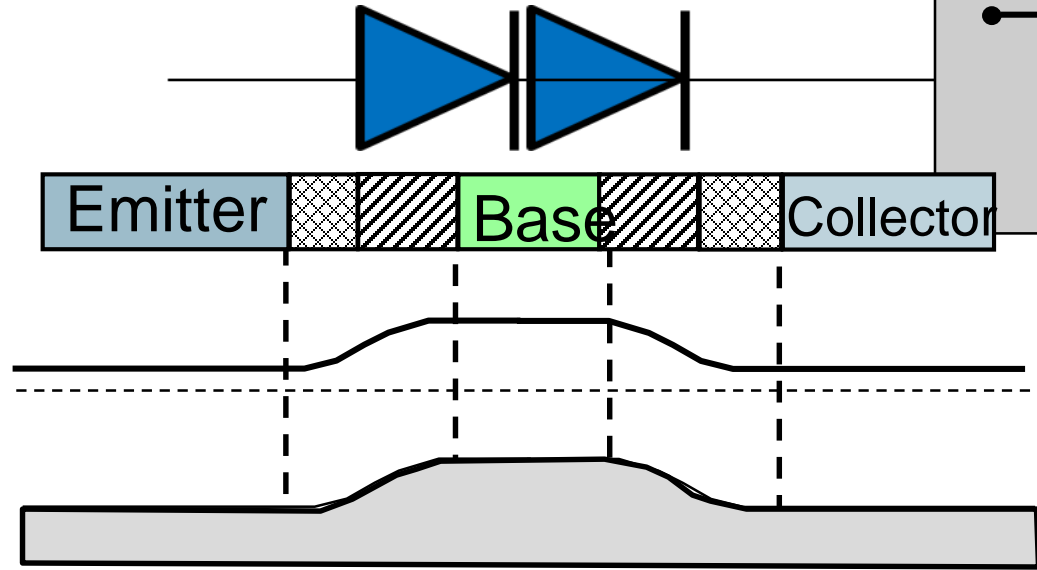
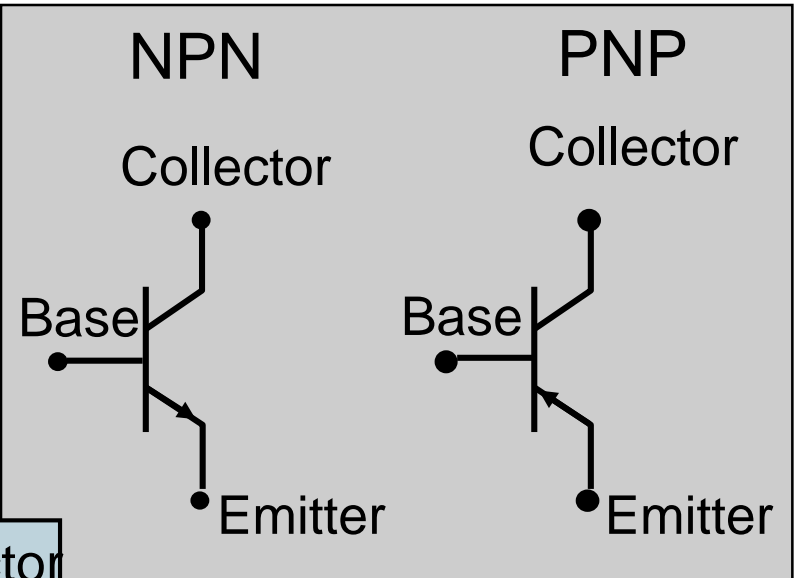
$$I = G \times V$$

$$= q \times n \times v \times A$$

↑ charge density
 ↑ velocity
 area

- >
- >
- >
- >

- 24.1 Introduction
- 24.2 Band Diagram in Equilibrium
- 24.3 Currents in BJTs
- 24.4 Ebers Moll Model



Band Diagram with Bias

$$\nabla \cdot D = q(p - n + N_D^+ - N_A^-)$$

$$\frac{\partial n}{\partial t} = \frac{1}{q} \nabla \cdot \mathbf{J}_N - r_N + g_N$$

$$\mathbf{J}_N = qn\mu_N E + qD_N \nabla n$$

$$\frac{\partial p}{\partial t} = \frac{1}{q} \nabla \cdot \mathbf{J}_P - r_P + g_P$$

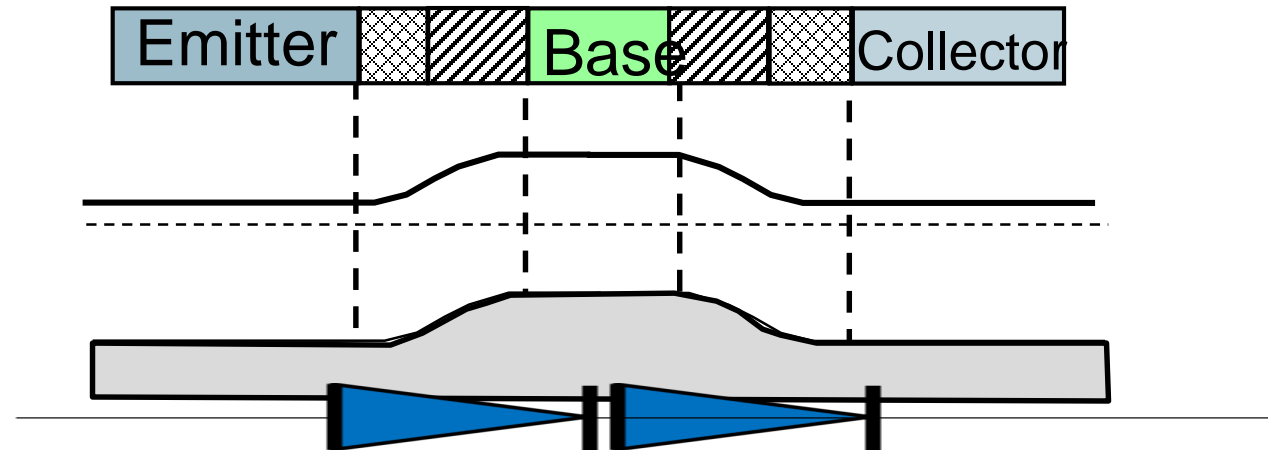
$$\mathbf{J}_P = qp\mu_P E - qD_P \nabla p$$

← **Non-equilibrium**

DC $dn/dt=0$

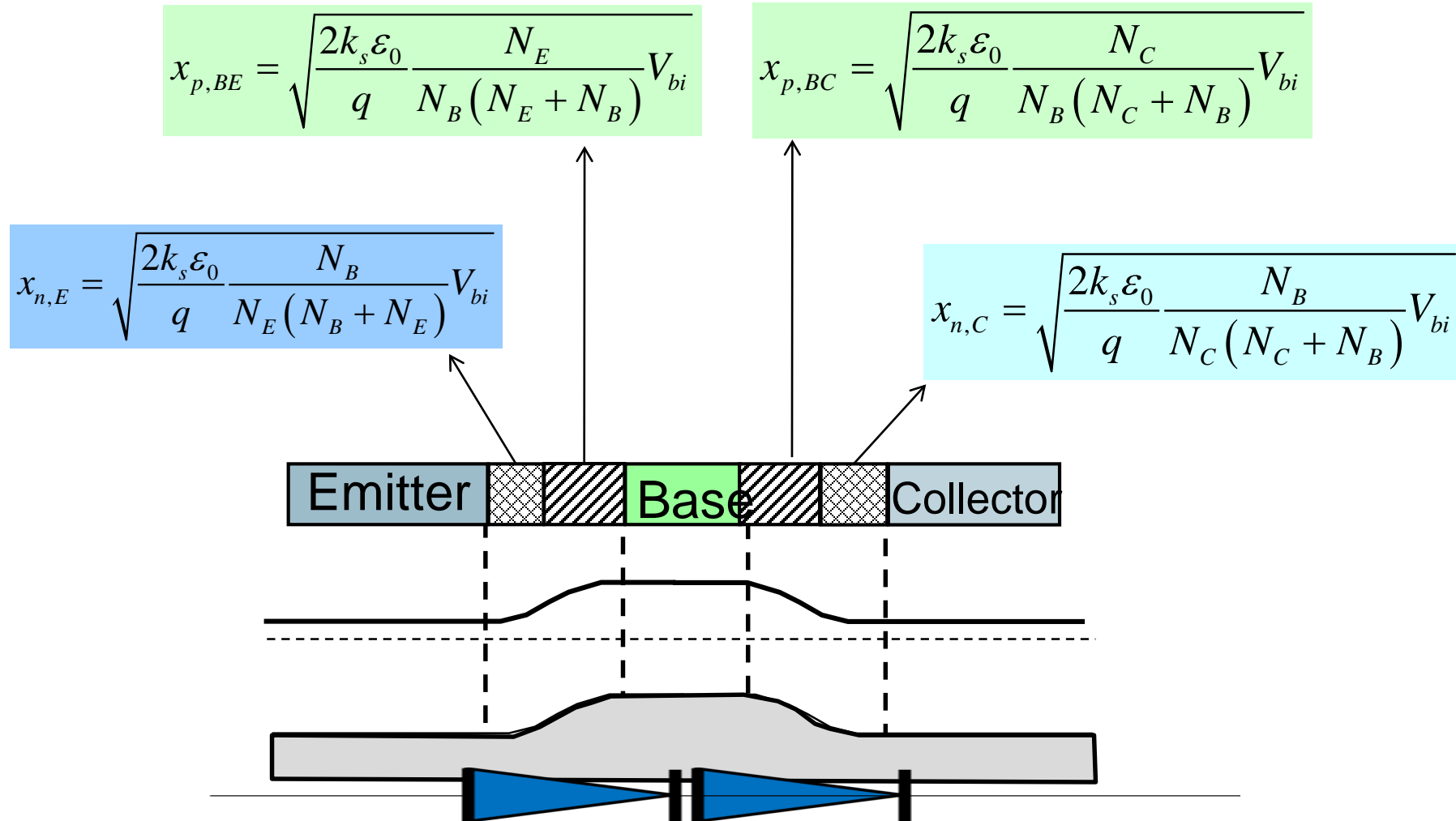
Small signal $dn/dt \sim j\omega n$

Transient --- Charge control model



Electrostatics in Equilibrium

Two back to back p-n junctions



Electrostatics in Equilibrium

$$x_{p,BE} = \sqrt{\frac{2k_s \epsilon_0}{q} \frac{N_E}{N_B (N_E + N_B)} (V_{bi} - V_{EB})}$$

$$x_{p,BC} = \sqrt{\frac{2k_s \epsilon_0}{q} \frac{N_C}{N_B (N_C + N_B)} (V_{bi} - V_{CB})}$$

$$x_{n,E} = \sqrt{\frac{2k_s \epsilon_0}{q} \frac{N_B}{N_E (N_B + N_E)} (V_{bi} - V_{EB})}$$

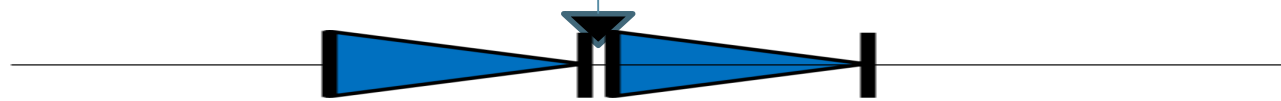
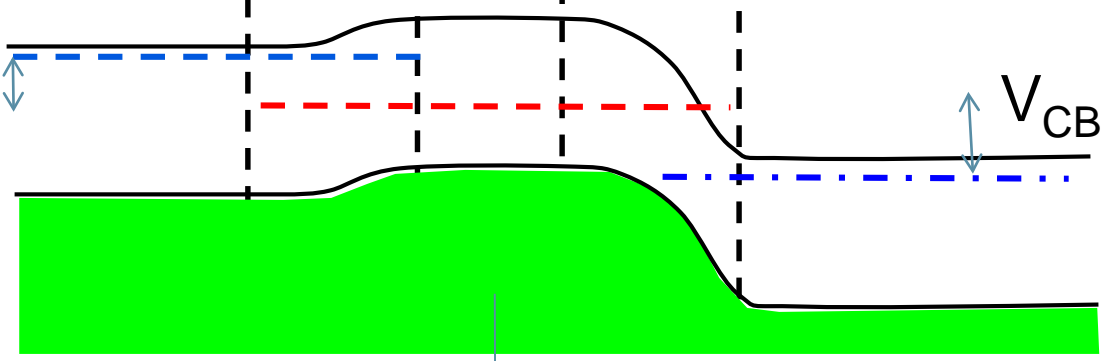
$$x_{n,C} = \sqrt{\frac{2k_s \epsilon_0}{q} \frac{N_B}{N_C (N_C + N_B)} (V_{bi} - V_{CB})}$$



V_{EB}

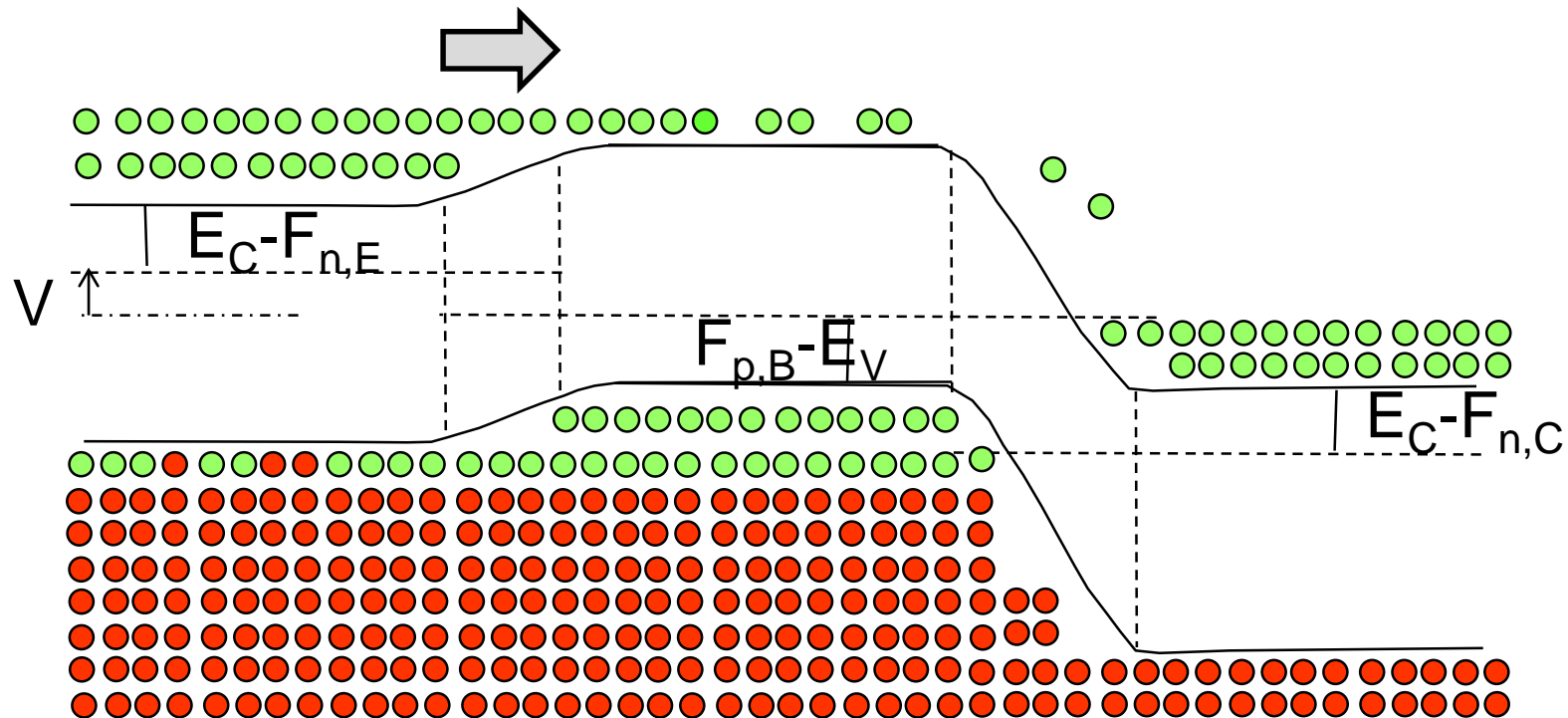
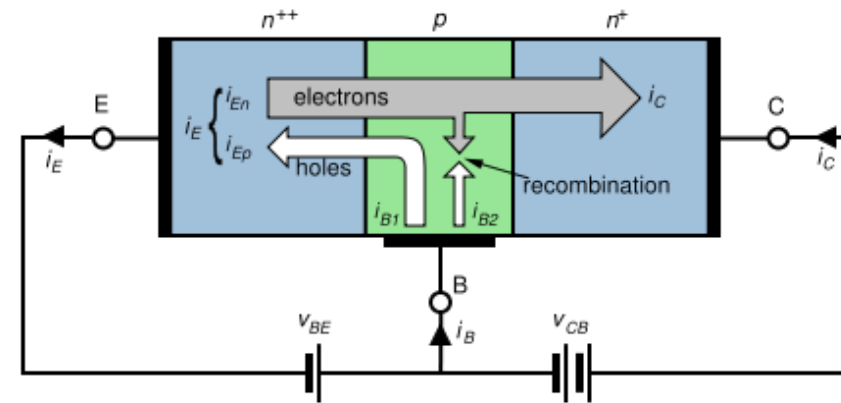
V_{CB}

Assume current flow is small...
Fermi level is flat



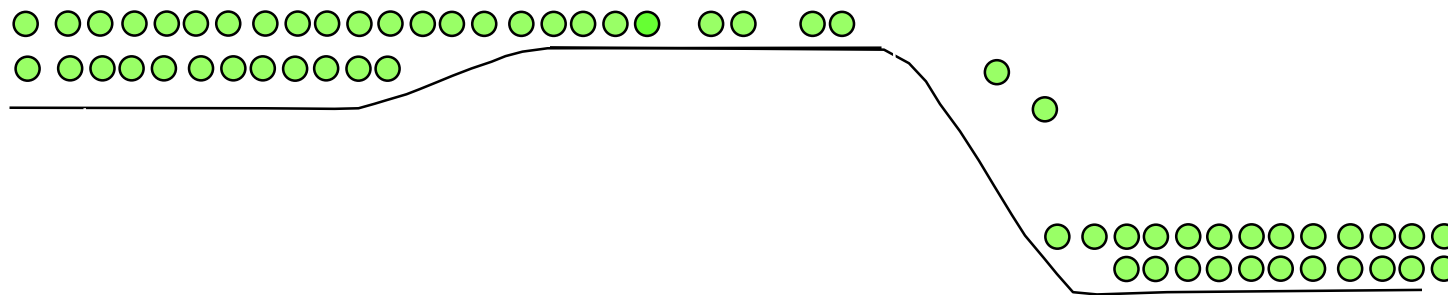
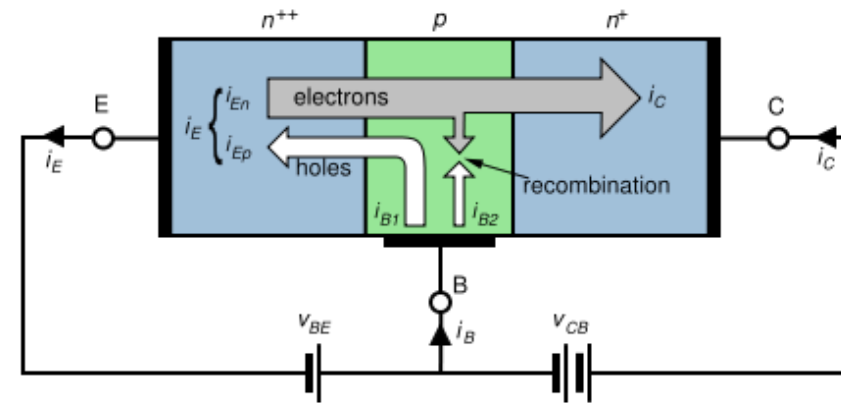
Current flow with Bias

Input small amount of holes results in large amount of electron output

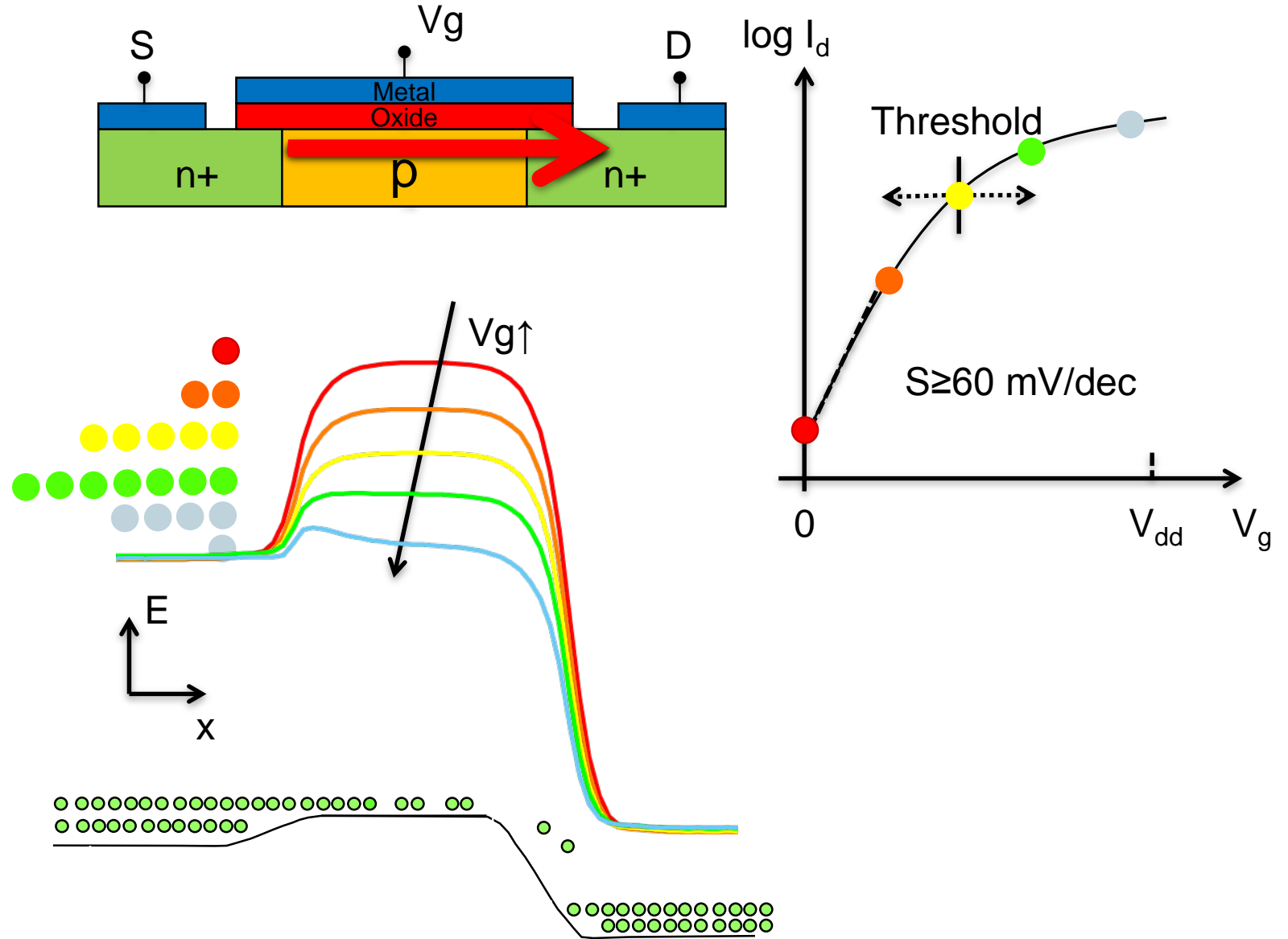


Current flow with Bias

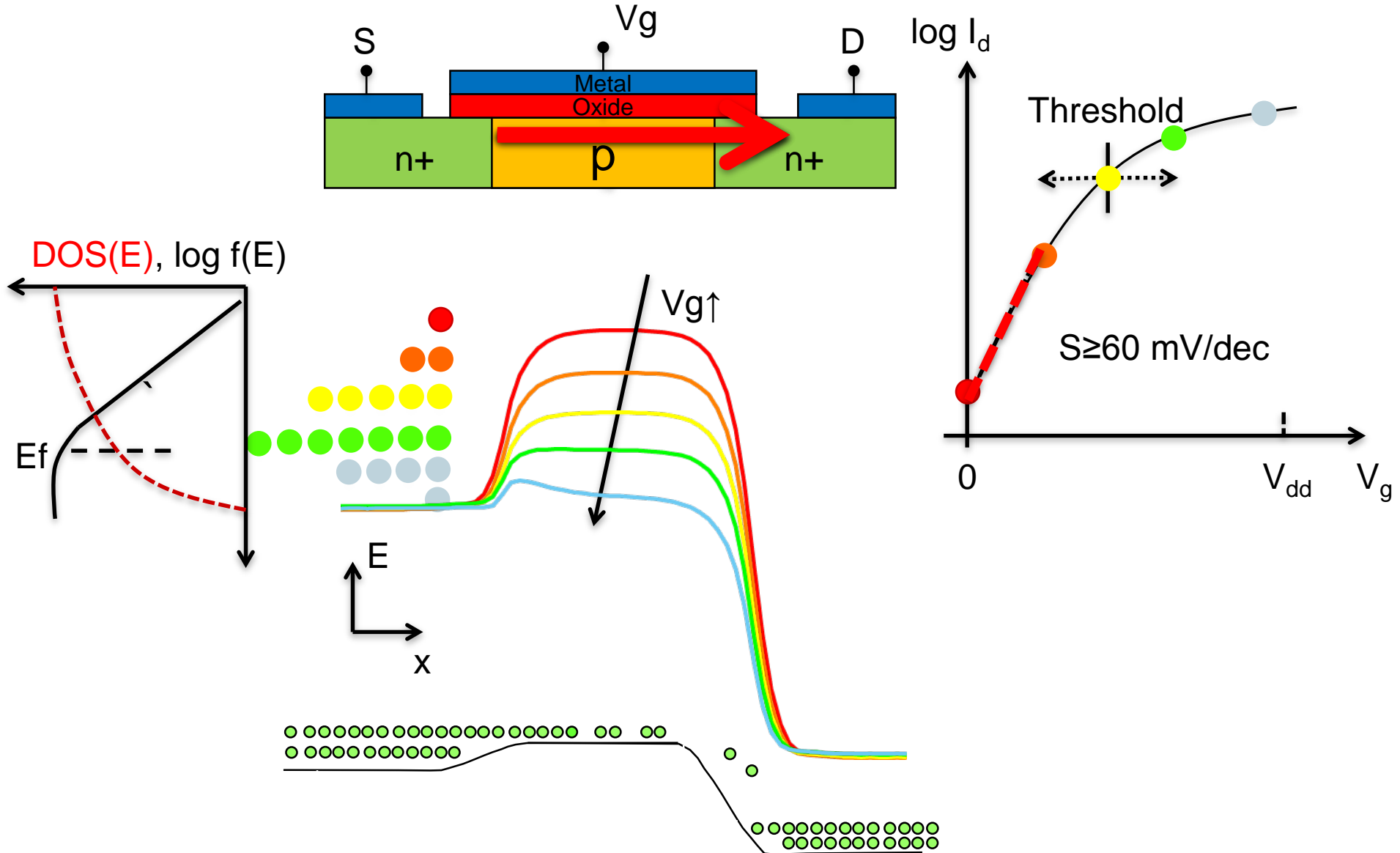
Input small amount of holes results in large amount of electron output



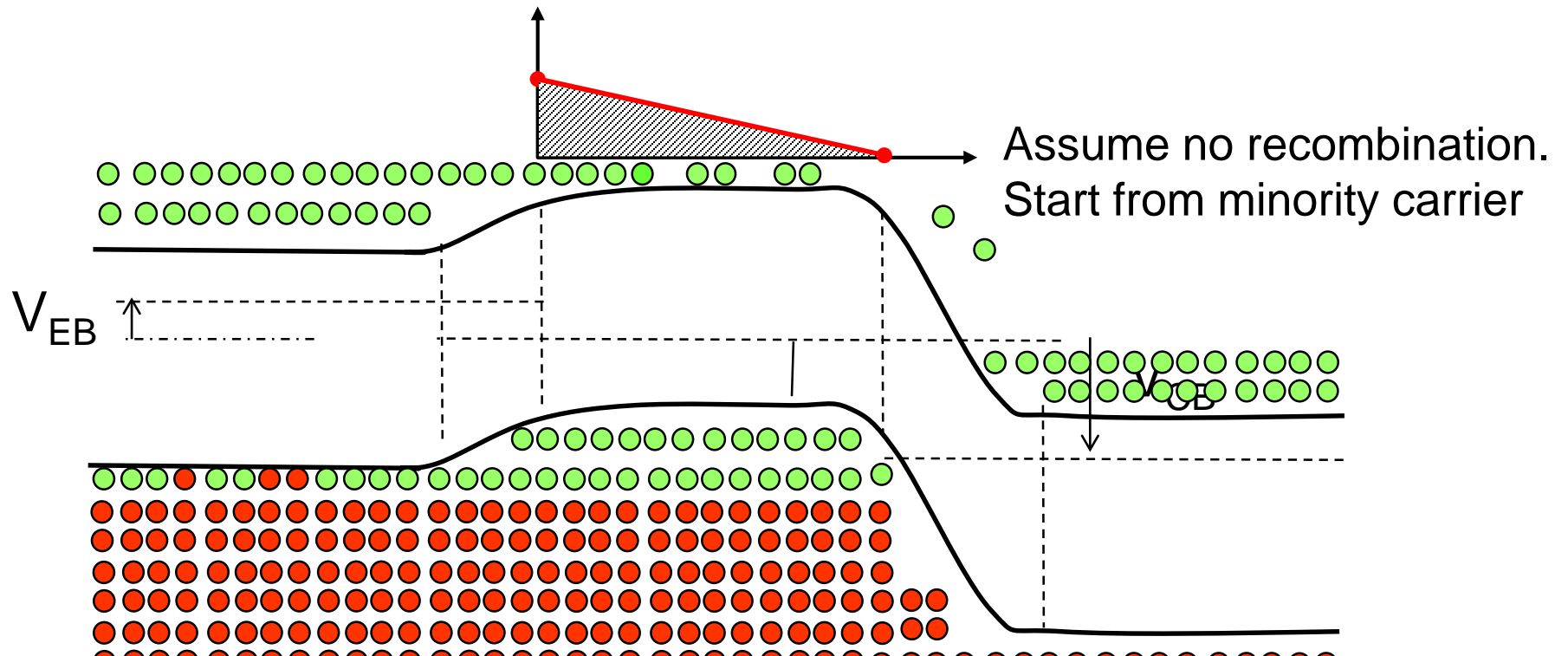
Modern MOSFET - "Fundamental" Limit looks similar to BJT



Modern MOSFET - "Fundamental" Limit looks similar to BJT



Carrier Distribution in Base



Carrier Distribution in Base

$$\Delta n(x) = Ax + B = C \left(1 - \frac{x}{W_B} \right) + D \left(\frac{x}{W_B} \right)$$

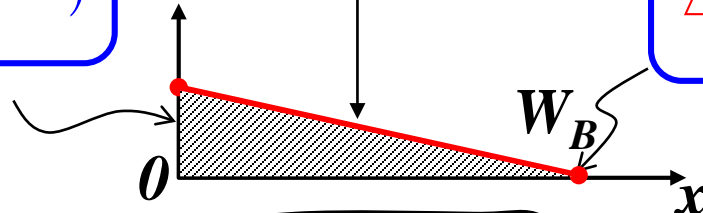
C

D

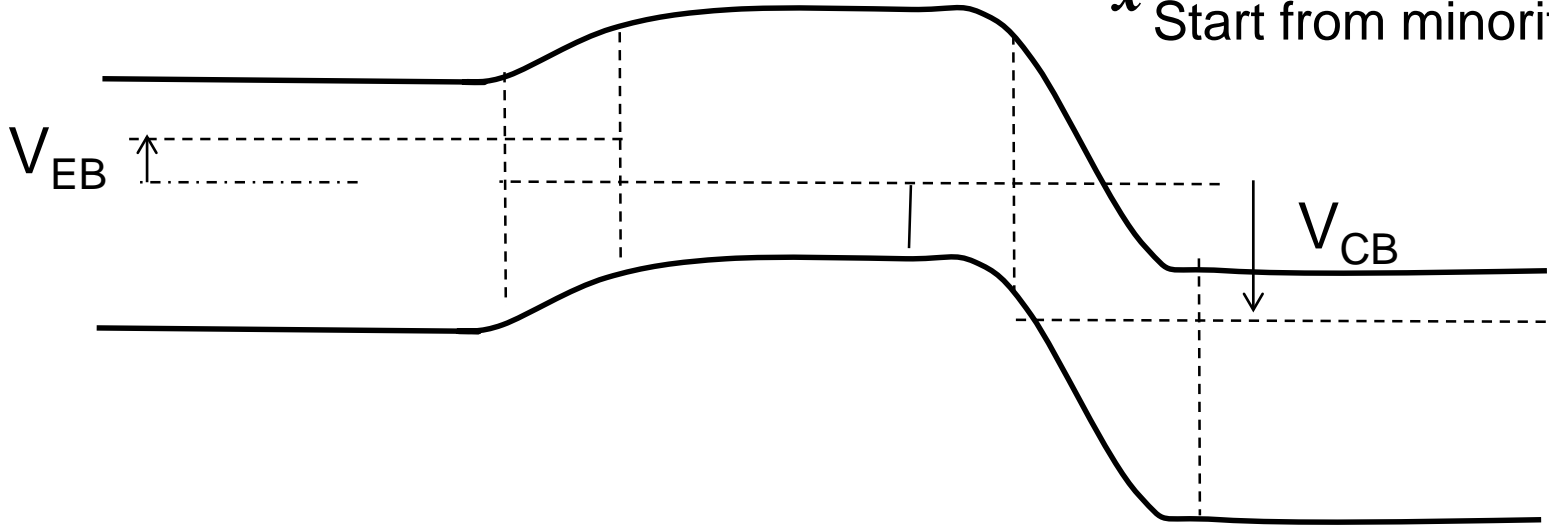
$$\Delta n(x) = \frac{n_{i,B}^2}{N_B} \left(e^{qV_{BE}\beta} - 1 \right) \left(1 - \frac{x}{W_B} \right) + \frac{n_{i,B}^2}{N_B} \left(e^{qV_{BC}\beta} - 1 \right) \left(\frac{x}{W_B} \right)$$

$$\Delta n(0^+) = \frac{n_{i,B}^2}{N_B} \left(e^{qV_{BE}\beta} - 1 \right)$$

$$\Delta n(x = W_B) = \frac{n_{i,B}^2}{N_B} \left(e^{qV_{BC}\beta} - 1 \right)$$



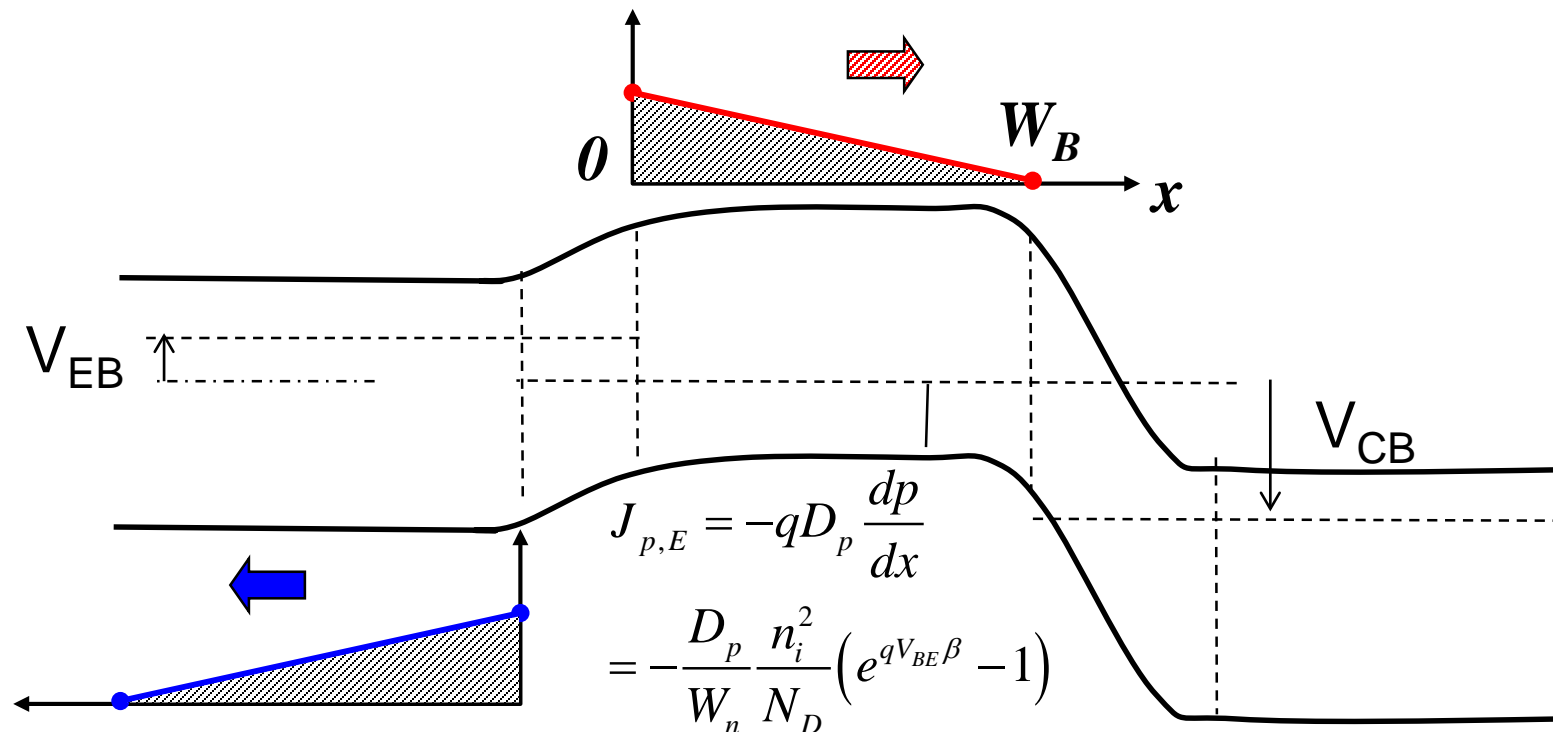
Assume no recombination.
Start from minority carrier



Collector and Emitter Electron Current

$$\Delta n(x) = \frac{n_{i,B}^2}{N_B} \left(e^{qV_{BE}\beta} - 1 \right) \left(1 - \frac{x}{W_B} \right) + \frac{n_{i,B}^2}{N_B} \left(e^{qV_{BC}\beta} - 1 \right) \left(\frac{x}{W_B} \right)$$

$$J_{n,C} = qD_n \left. \frac{dn}{dx} \right|_{W_B} \quad J_{n,C} = -\frac{qD_n n_{i,B}^2}{W_B N_B} \left(e^{qV_{BE}\beta} - 1 \right) + \frac{qD_n n_{i,B}^2}{W_B N_B} \left(e^{qV_{BC}\beta} - 1 \right)$$



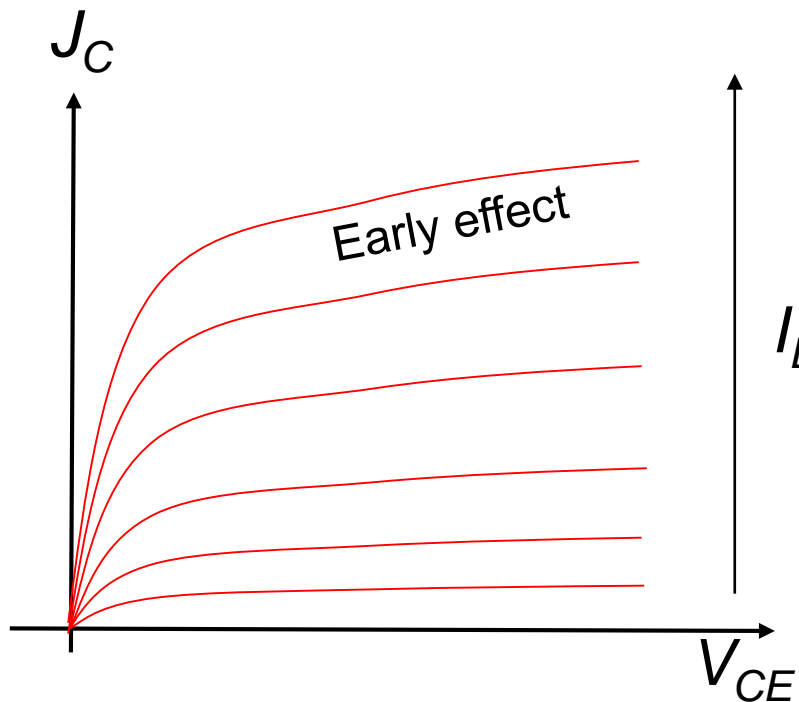
Current-Voltage Characteristics

Normal, Active Region

EB: Forward biased

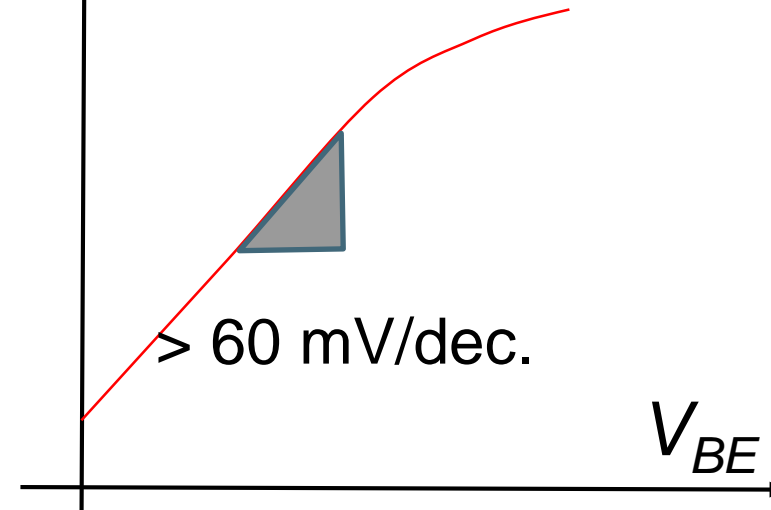
BC: Reverse biased

$$J_{n,C} = -\frac{qD_n n_{i,B}^2}{W_B N_B} (e^{qV_{BE}\beta} - 1) + \frac{qD_n n_{i,B}^2}{W_B N_B} (e^{qV_{BC}\beta} - 1)$$



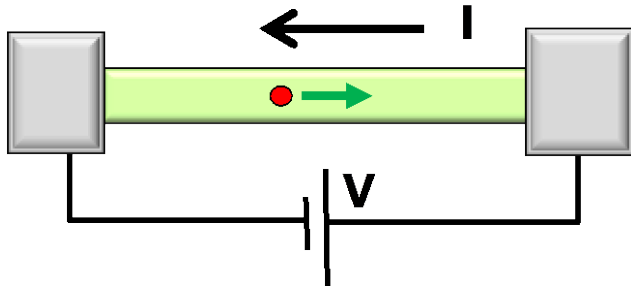
W_B is not independent of bias
 \Rightarrow Early Effect

$\log_{10} J_C$ High-level injection
series resistance, etc.



same physics of diode, rollover

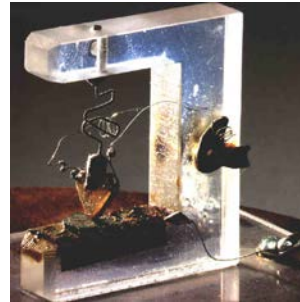
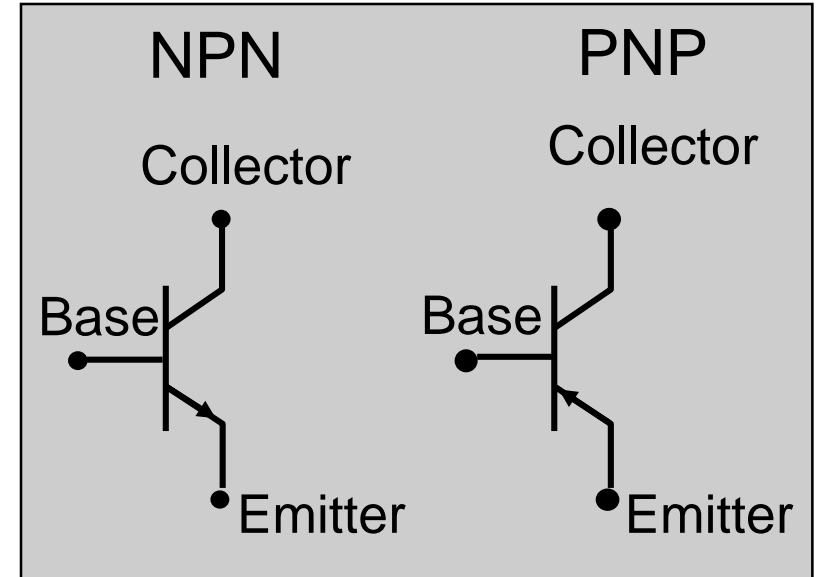
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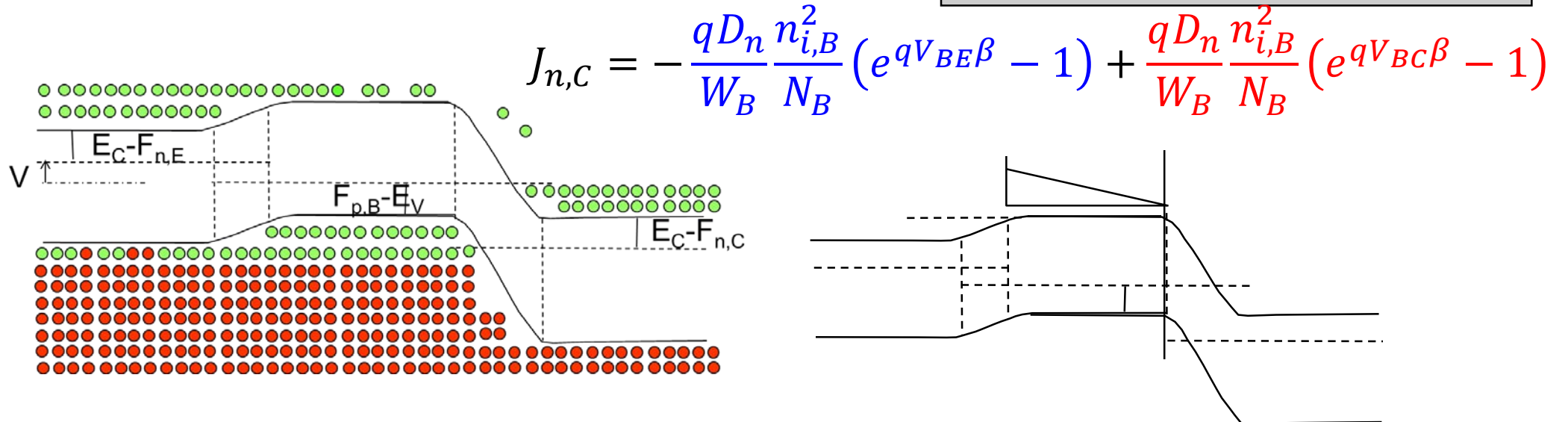
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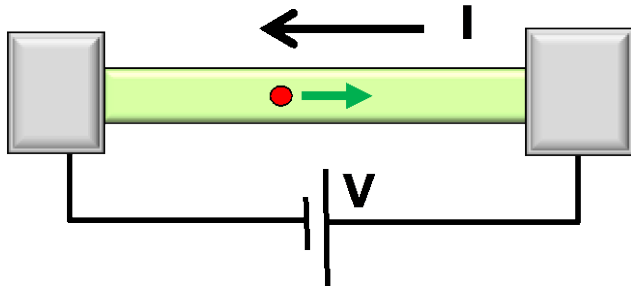
↑ charge density
 ↑ velocity
 area



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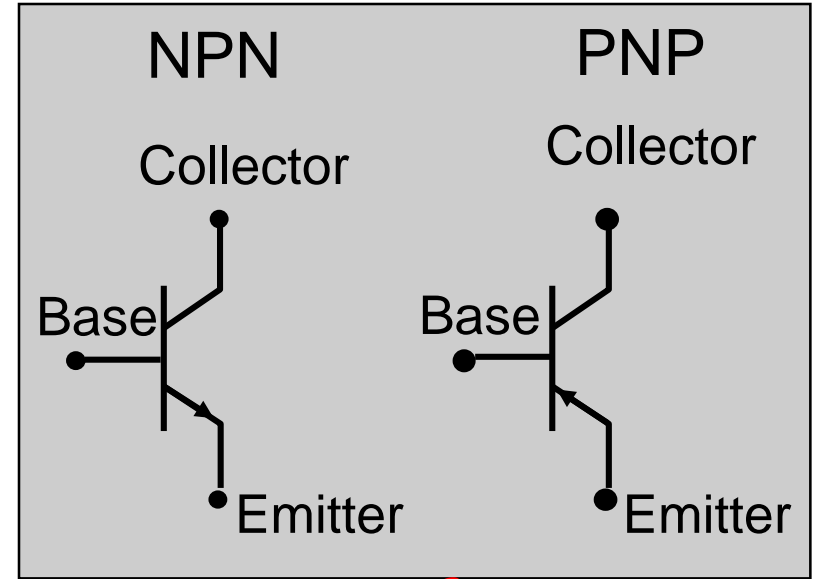
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