

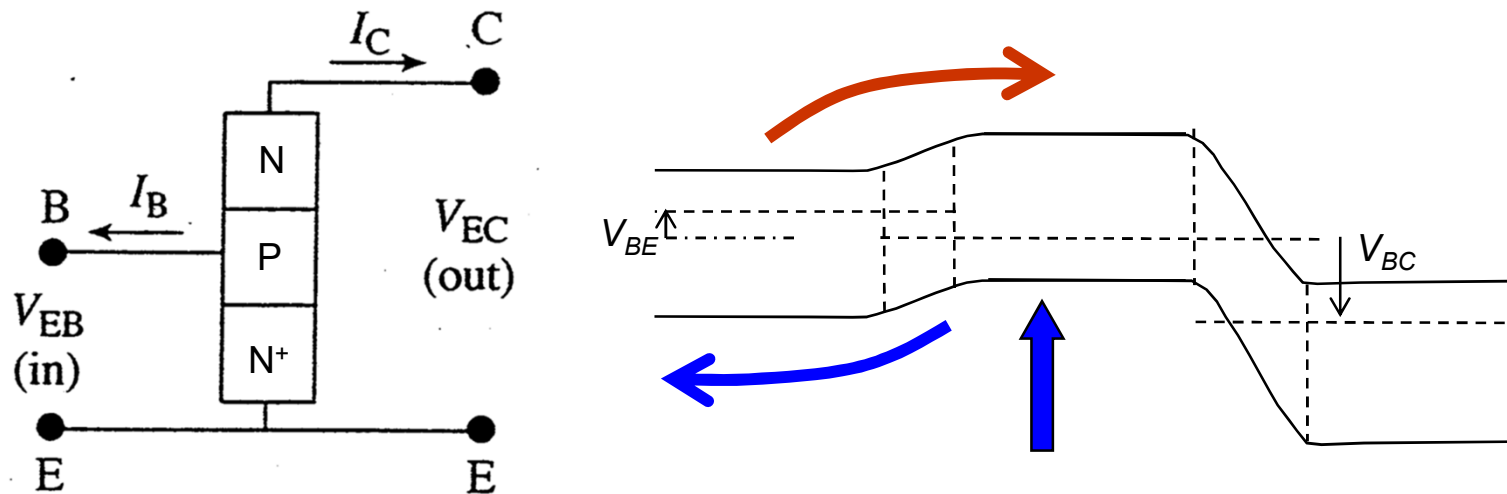
# **ECE-305: Spring 2018**

## **Bipolar Junction Transistor I-V**

Pierret, *Semiconductor Device Fundamentals* (SDF)  
Chapters 10 and 11 (pp. 371-385, 389-403)

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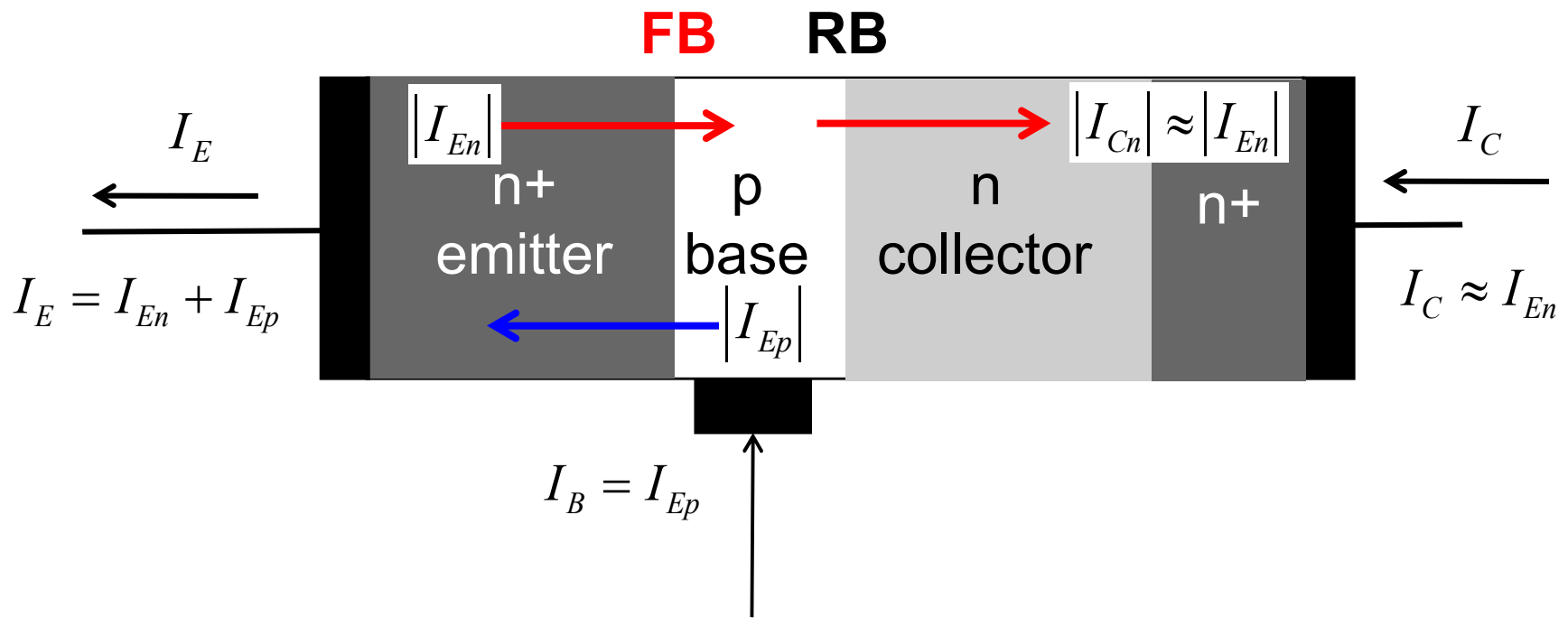
# essence of current gain



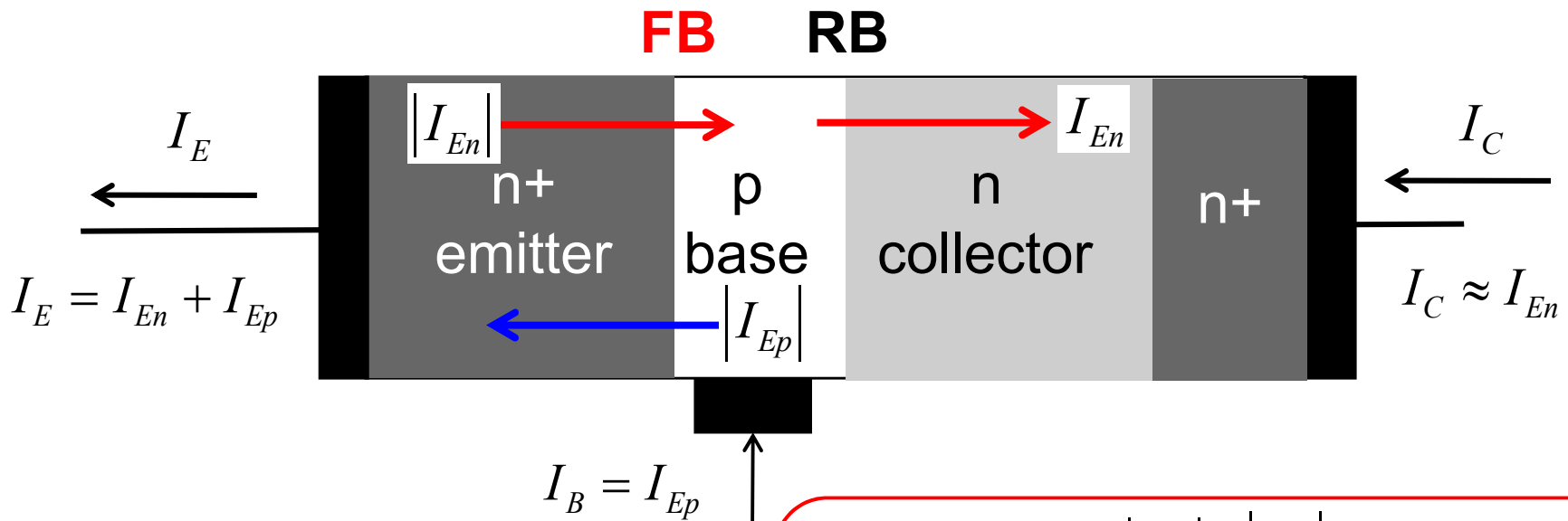
Input  $\downarrow$   $I_B \approx \frac{qD_p}{W_E} \frac{n_{i,E}^2}{N_E} (e^{qV_{BE}\beta} - 1)$   $\uparrow$  Response

Response  $\leftarrow$  Input  $\downarrow$   $I_E \approx \frac{qD_n}{W_B} \frac{n_{i,B}^2}{N_B} (e^{qV_{BE}\beta} - 1)$   $\downarrow$  Response

# forward active region



# emitter current: forward active region



$$|I_{En}| = qA_E \left( \frac{n_i^2}{N_{AB}} \right) \frac{D_n}{W_B} e^{qV_{BE}/k_B T}$$

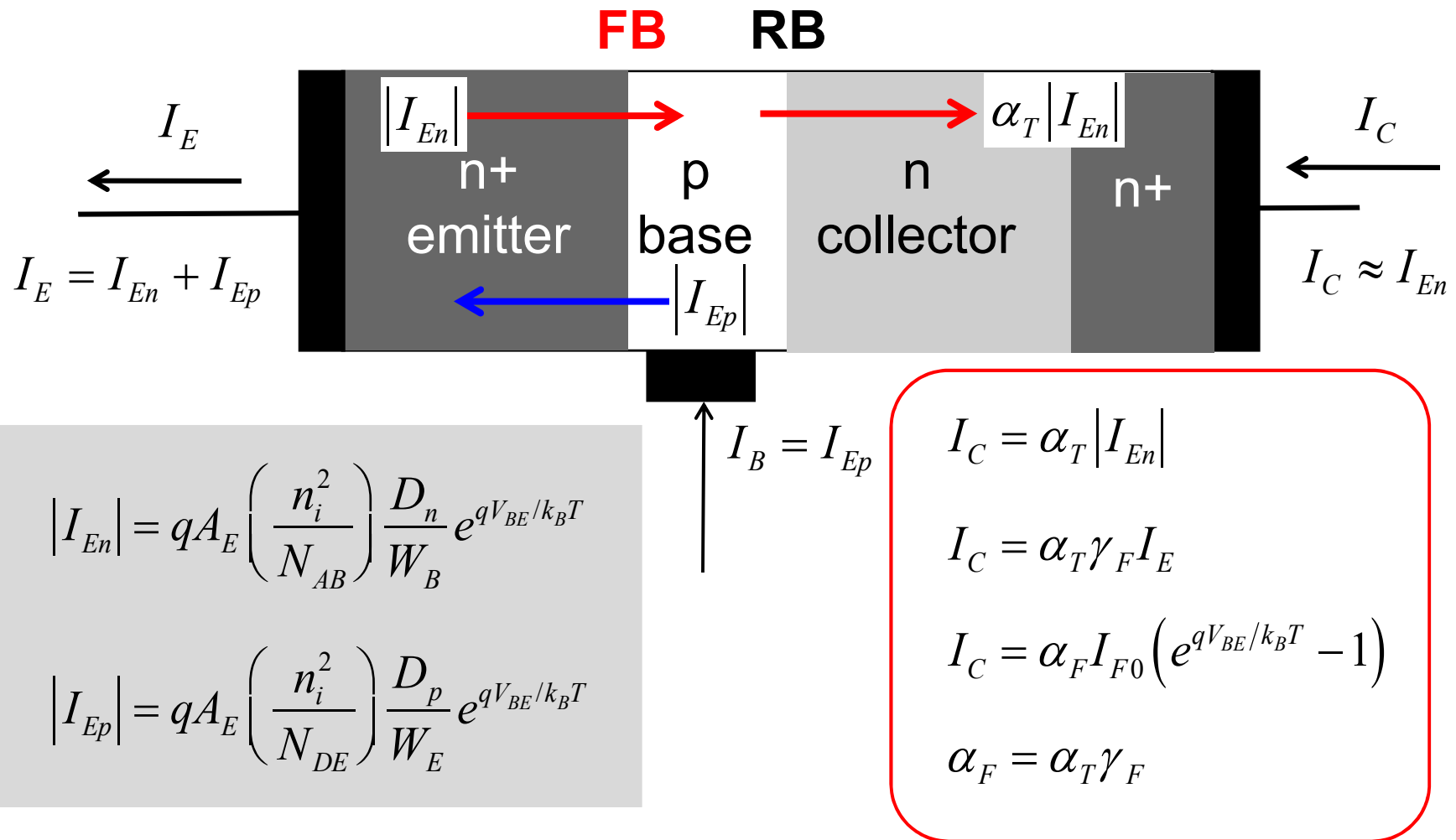
$$|I_{Ep}| = qA_E \left( \frac{n_i^2}{N_{DE}} \right) \frac{D_p}{W_E} e^{qV_{BE}/k_B T}$$

$$I_E = |I_{En}| + |I_{Ep}|$$

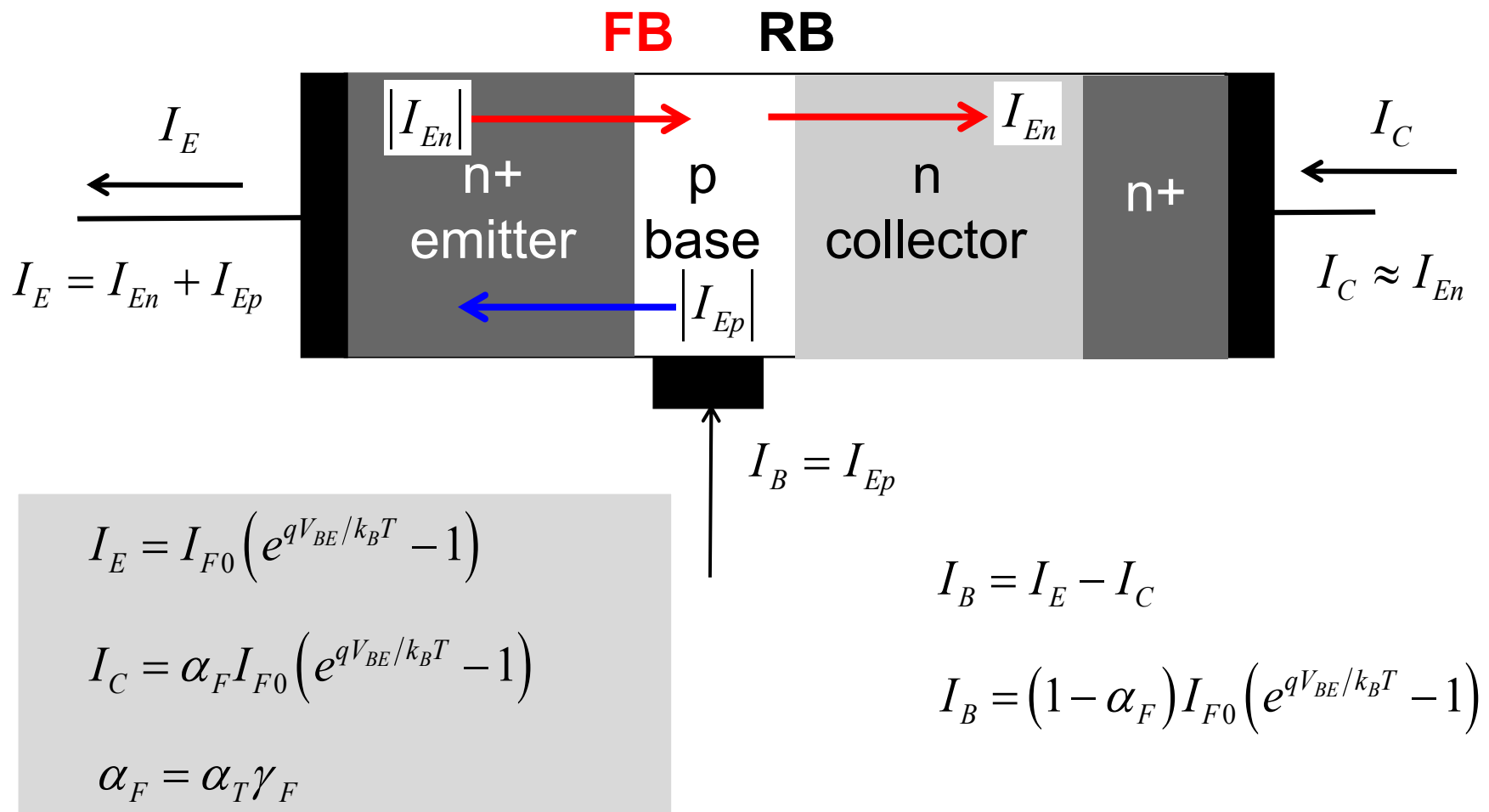
$$I_E = I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right)$$

$$I_{F0} = qA_E \left( \frac{n_i^2}{N_{AB}} \right) \frac{D_n}{W_B} + qA_E \left( \frac{n_i^2}{N_{DE}} \right) \frac{D_p}{W_E}$$

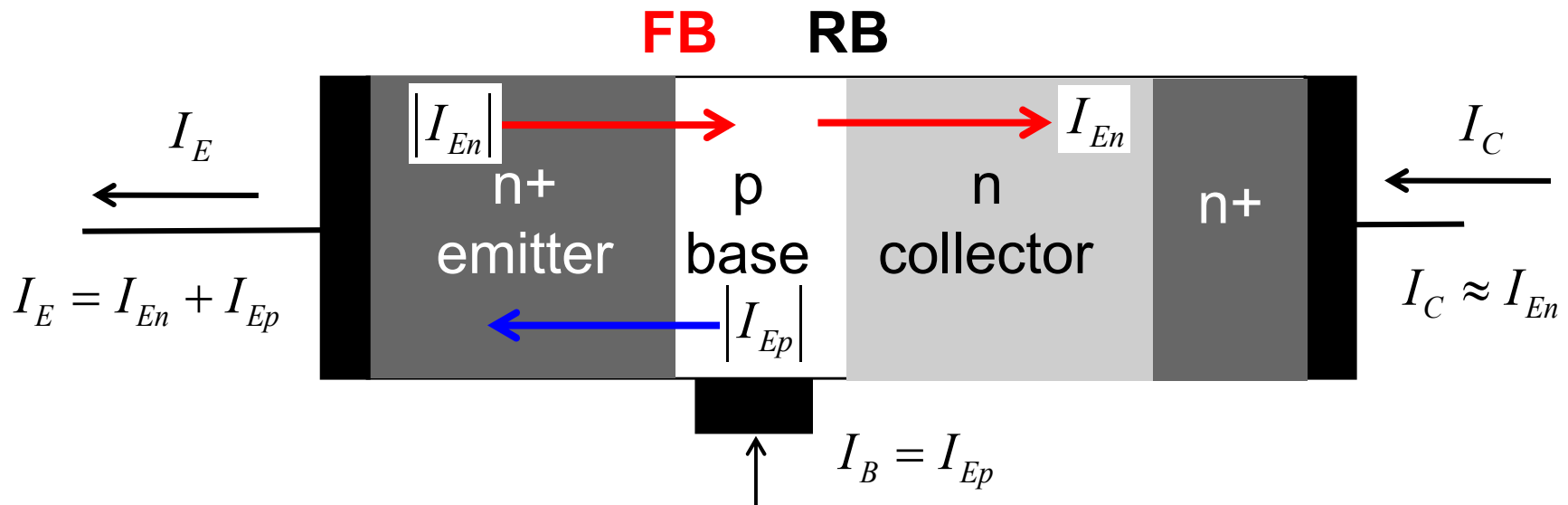
# collector current: forward active region



# base current: forward active region



# summary: forward active region



$$I_E = I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right)$$

$$\alpha_F = \alpha_T \gamma_F$$

$$I_C = \alpha_F I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right)$$

$$I_{F0} = qA_E \left( \frac{n_i^2}{N_{AB}} \right) \frac{D_n}{W_B} + qA_E \left( \frac{n_i^2}{N_{DE}} \right) \frac{D_p}{W_E}$$

$$I_B = (1 - \alpha_F) I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right)$$

# Ebers-Moll model

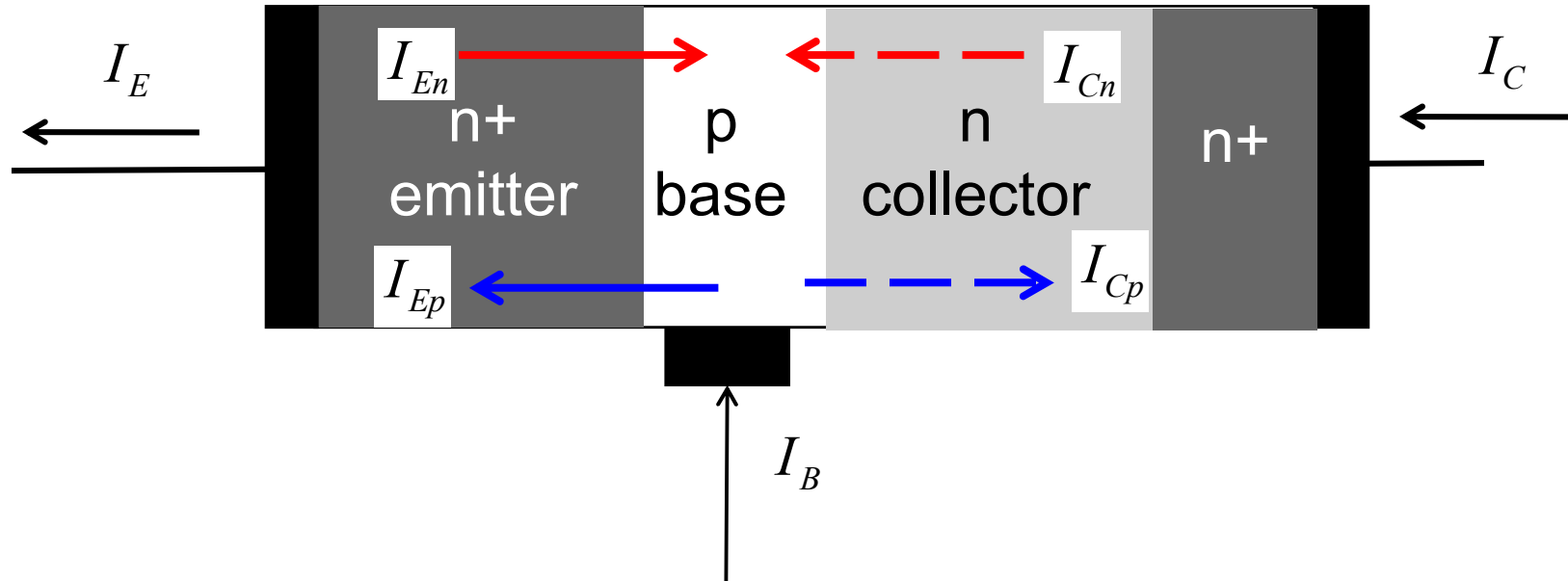
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Question:

How do we describe the BJT in **any** region of operation?



# emitter-base junction (the **forward diode**)



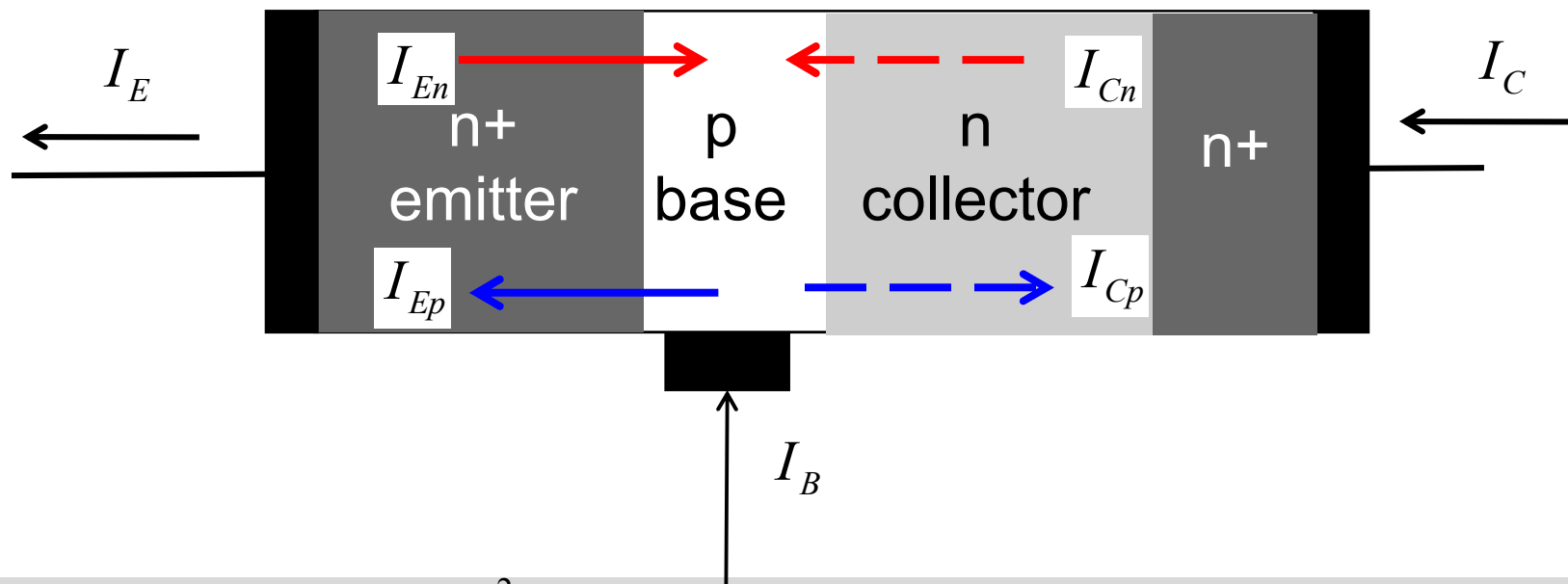
$$I_{En} = -qA \frac{D_n}{W_B} \frac{n_i^2}{N_{AB}} \left( e^{qV_{BE}/k_B T} - 1 \right)$$

$$I_E(V_{BE}) = -I_{En}(V_{BE}) - I_{Ep}(V_{BE})$$

$$I_{Ep} = -qA \frac{D_p}{W_E} \frac{n_i^2}{N_{DE}} \left( e^{qV_{BE}/k_B T} - 1 \right)$$

$$I_E(V_{BE}) = I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right)$$

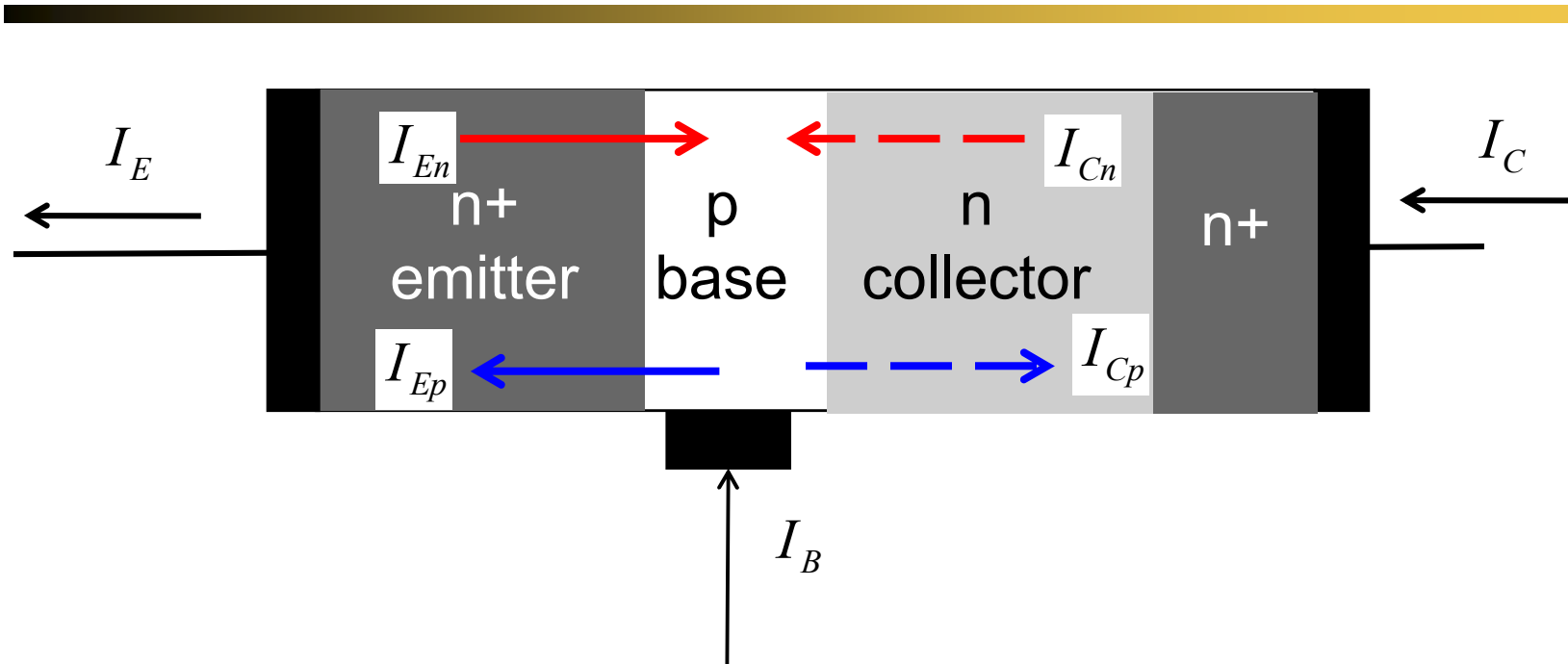
# Base-collector junction (the reverse diode)



$$I_{Cn}(V_{BC}) = qA \frac{D_n}{W_B} \frac{n_i^2}{N_{AB}} \left( e^{qV_{BC}/k_B T} - 1 \right) \quad I_C(V_{BC}) = - \left[ I_{Cn}(V_{BC}) + I_{Cp}(V_{BC}) \right]$$

$$I_{Cp}(V_{BC}) = qA \frac{D_p}{W_C} \frac{n_i^2}{N_{DC}} \left( e^{qV_{BC}/k_B T} - 1 \right) \quad I_C(V_{BC}) = -I_{R0} \left( e^{qV_{BC}/k_B T} - 1 \right)$$

# Both junctions....



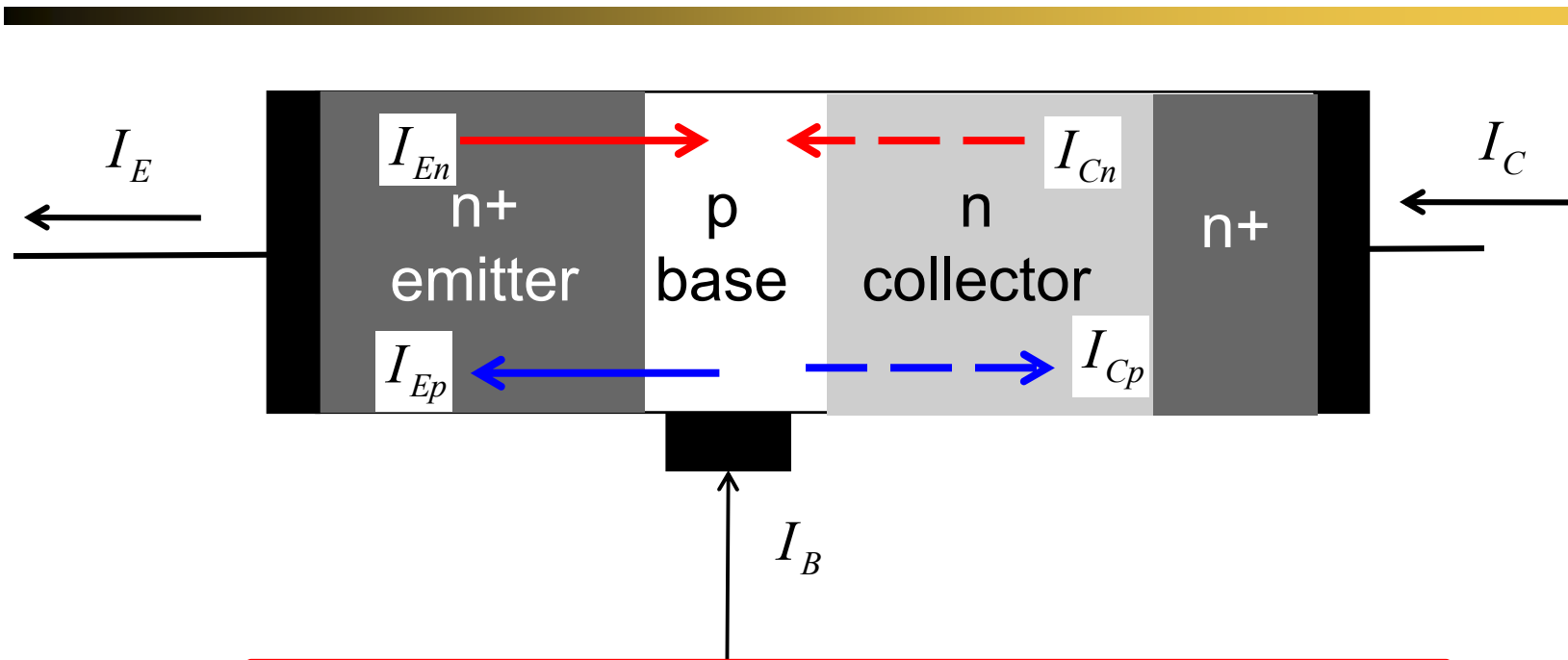
$$I_C(V_{BC}) = -I_{R0} \left( e^{qV_{BC}/k_B T} - 1 \right)$$

$$I_E(V_{BE}) = I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right)$$

**But....**

The two junctions are coupled!

# Ebers-Moll model



$$I_C(V_{BE}, V_{BC}) = \alpha_F I_{F0} (e^{qV_{BE}/k_B T} - 1) - I_{R0} (e^{qV_{BC}/k_B T} - 1)$$

$$I_E(V_{BE}, V_{BC}) = I_{F0} (e^{qV_{BE}/k_B T} - 1) - \alpha_R I_{R0} (e^{qV_{BC}/k_B T} - 1)$$

# Ebers-Moll model

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$$I_C(V_{BE}, V_{BC}) = \alpha_F I_{F0} (e^{qV_{BE}/k_B T} - 1) - I_{R0} (e^{qV_{BC}/k_B T} - 1)$$

$$I_E(V_{BE}, V_{BC}) = I_{F0} (e^{qV_{BE}/k_B T} - 1) - \alpha_R I_{R0} (e^{qV_{BC}/k_B T} - 1)$$

$$I_B(V_{BE}, V_{BC}) = I_E(V_{BE}, V_{BC}) - I_C(V_{BE}, V_{BC})$$

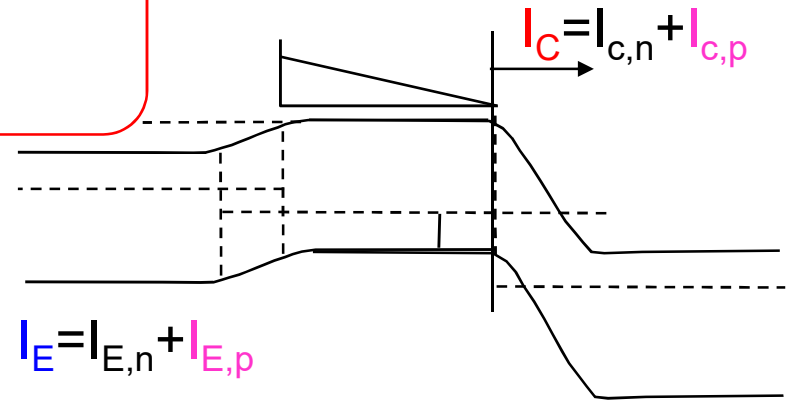
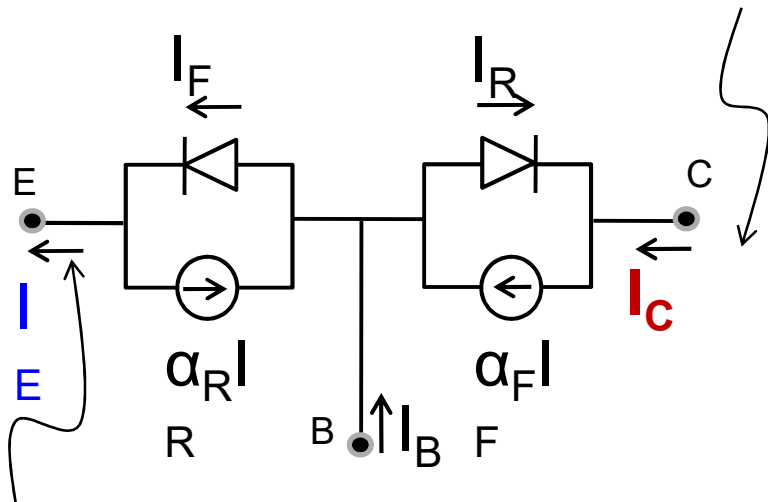
$$\alpha_F I_{F0} = \alpha_R I_{R0}$$

See Pierret SDF, Chapter 11, sec. 11.1.4

# Ebers Moll Model

$$I_C = -A \frac{qD_n n_{i,B}^2}{W_B N_B} (e^{qV_{BE}\beta} - 1) + A \left[ \frac{qD_n n_{i,B}^2}{W_B N_B} + \frac{qD_p n_{i,C}^2}{W_C N_C} \right] (e^{qV_{BC}\beta} - 1)$$

$$\equiv \alpha_F I_{F0} (e^{qV_{BE}\beta} - 1) - I_{R0} (e^{qV_{BC}\beta} - 1)$$



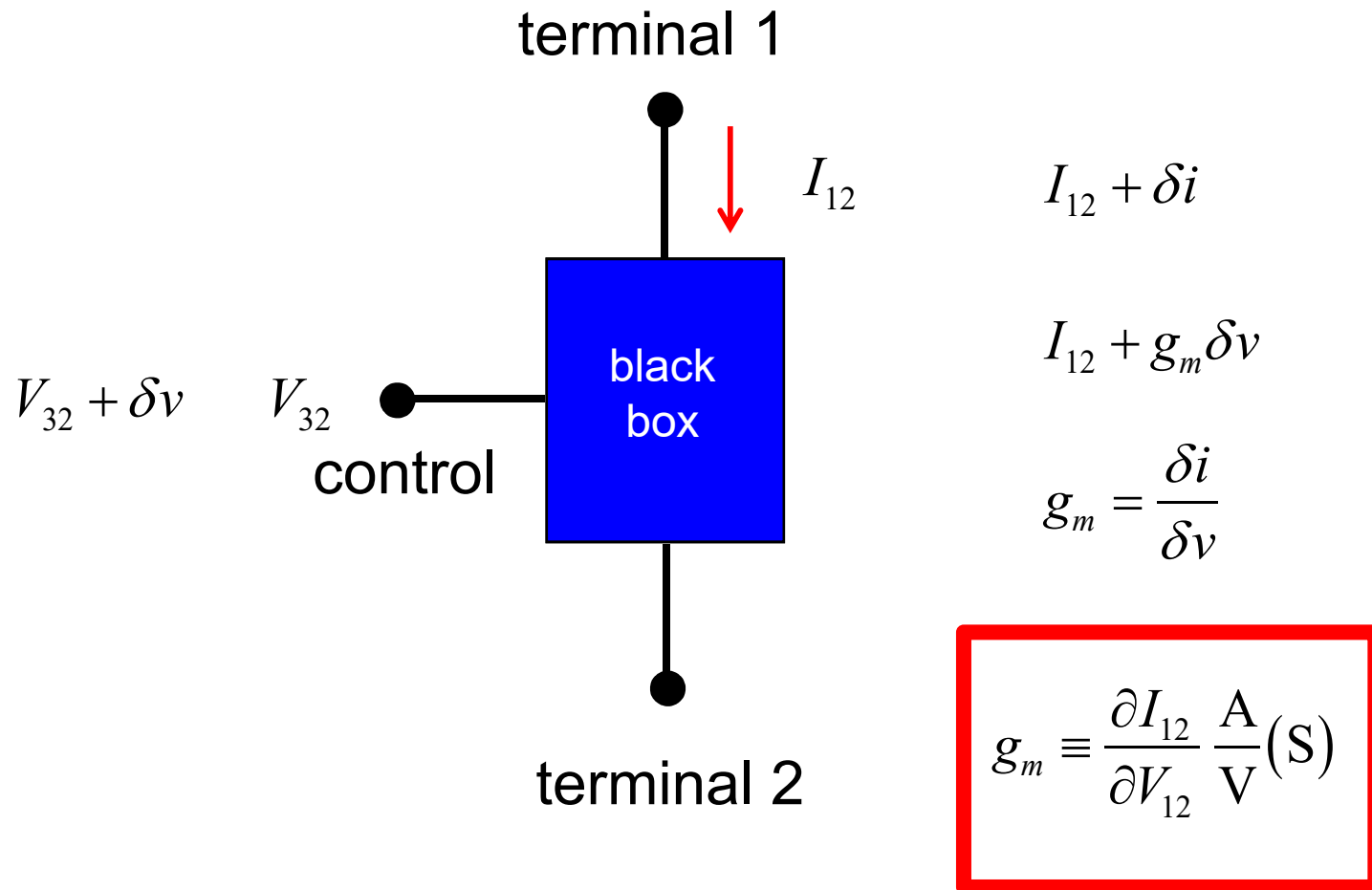
$$I_F = I_{F0} (e^{qV_{BE}\beta} - 1)$$

$$I_R = I_{R0} (e^{qV_{BC}\beta} - 1)$$

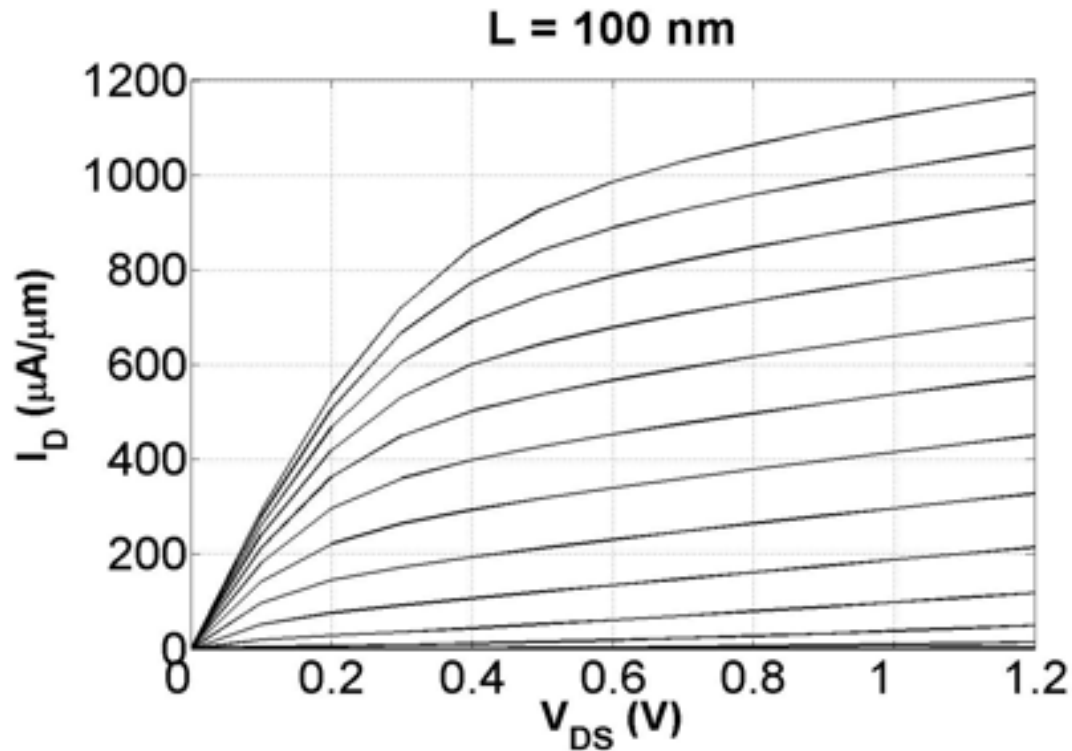
$$I_E = -A \left[ \frac{qD_p n_{i,E}^2}{W_E N_E} + \frac{qD_n n_{i,B}^2}{W_B N_B} \right] (e^{qV_{BE}\beta} - 1) + A \frac{qD_n n_{i,B}^2}{W_E N_B} (e^{qV_{BC}\beta} - 1)$$

$$\equiv I_{F0} (e^{qV_{BE}\beta} - 1) - \alpha_R I_{R0} (e^{qV_{BC}\beta} - 1)$$

# transconductance



# transconductance



$$V_{GS} = 1.2 \text{ V}$$

$$V_{GS} = 1.1 \text{ V}$$

$$g_m = \left. \frac{\Delta I_D}{\Delta V_{GS}} \right|_{V_{DS}} \frac{\text{A}}{\text{V}} = \text{S}$$

$$g_m = 1000 \text{ mS/mm}$$

$$g_m = \frac{(1180 - 1080) \times 10^{-6}}{0.1} = 0.001 \text{ S}/\mu\text{m}$$

$$g_m = 1000 \mu\text{S}/\mu\text{m}$$



# MOSFETs vs. BJTs

## **MOS (saturated)**

$$g_m = \partial I_D / \partial V_{GS} \Big|_{V_{DS}}$$

$$I_D = WC_{ox} \langle v(0) \rangle (V_{GS} - V_T)$$

$$g_m = WC_{ox} \langle v(0) \rangle$$

$$g_m / I_D = 1 / (V_{GS} - V_T)$$

$$g_m / I_D = 1 / (1.0 - 0.2) \approx 1.25 \text{ V}^{-1}$$

## **Bipolar (active)**

$$g_m = \partial I_C / \partial V_{BE} \Big|_{V_{CE}}$$

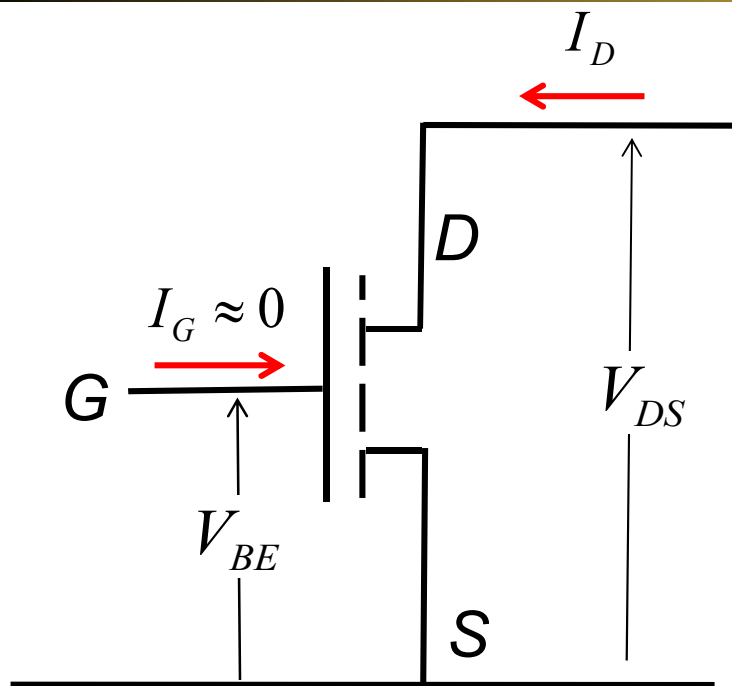
$$I_C = \alpha_F I_{F0} e^{qV_{BE}/k_B T}$$

$$g_m = I_C / (k_B T / q)$$

$$g_m / I_C = 1 / (k_B T / q)$$

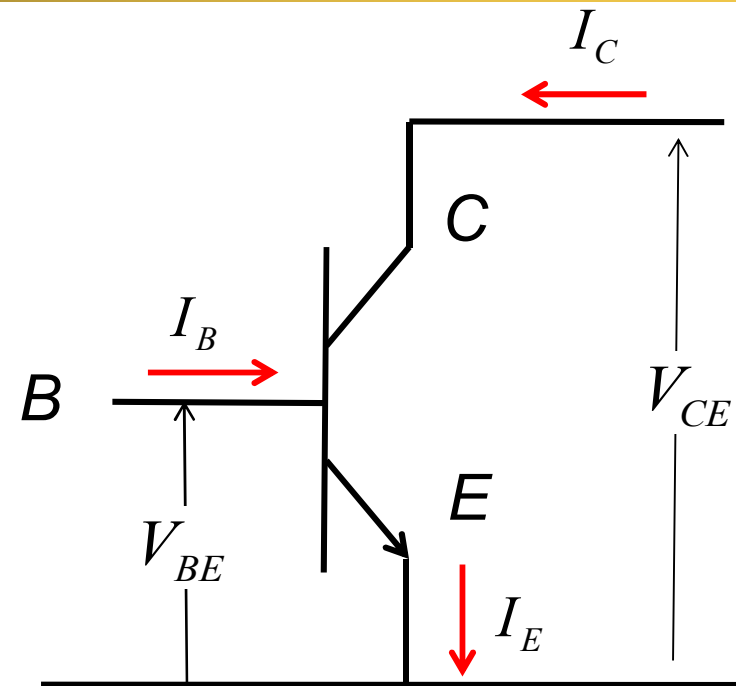
$$g_m / I_C = 1 / (0.026) \approx 40 \text{ V}^{-1}$$

# MOSFETs vs. BJTs



MOSFET characteristics:

- simple to make
- no gate current
- moderate  $g_m$
- low capacitance



BJT characteristics:

- more complex to make
- base current
- large  $g_m$
- high capacitance

# active region example 1 (NPN)

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$$A_E = 10 \mu\text{m} \times \mu\text{m}$$

$$I_C = 50 \mu\text{A}$$

$$N_{DE} = 1.5 \times 10^{19} \text{ cm}^{-3}$$

$$W_B = 0.50 \mu\text{m}$$

$$\tau_{pE} = 0.1 \text{ ns}$$

$$N_{AB} = 1.0 \times 10^{17} \text{ cm}^{-3}$$

$$W_B = 0.25 \mu\text{m}$$

$$\tau_{nB} = 75 \text{ ns}$$

$$N_{DC} = 2.0 \times 10^{16} \text{ cm}^{-3}$$

$$W_C = 1.50 \mu\text{m}$$

$$\tau_{pC} = 150 \text{ ns}$$

# Step 1: diffusion coefficients and lengths

$$N_{DE} = 1.5 \times 10^{19} \text{ cm}^{-3} \quad W_E = 0.50 \text{ } \mu\text{m} \quad \tau_{pE} = 0.1 \text{ ns}$$

$$N_{AB} = 1.0 \times 10^{17} \text{ cm}^{-3} \quad W_B = 0.25 \text{ } \mu\text{m} \quad \tau_{nB} = 75 \text{ ns}$$

$$N_{DC} = 2.0 \times 10^{16} \text{ cm}^{-3} \quad W_C = 1.50 \text{ } \mu\text{m} \quad \tau_{pC} = 150 \text{ ns}$$

Pierret, SDF, Fig. 3.5a, p. 80

$$\mu_{pE} \approx 70 \text{ cm}^2/\text{V-s} \quad D_{pE} = \frac{k_B T}{q} \mu_{pE} \approx 1.8 \text{ cm}^2/\text{s} \quad L_{pE} = \sqrt{D_{pE} \tau_{pE}} \approx 0.13 \text{ } \mu\text{m}$$

$$\mu_{nB} \approx 800 \text{ cm}^2/\text{V-s} \quad D_{nB} \approx 20.8 \text{ cm}^2/\text{s} \quad L_{nB} \approx 12.5 \text{ } \mu\text{m}$$

$$\mu_{pC} \approx 420 \text{ cm}^2/\text{V-s} \quad D_{pC} \approx 10.9 \text{ cm}^2/\text{s} \quad L_{pC} \approx 12.8 \text{ } \mu\text{m}$$

# find the emitter injection efficiency

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$$\gamma_F = \frac{1}{1 + \frac{D_{pE} W_B N_{AB}}{D_{nB} W_E N_{DE}}}$$

$$\gamma_F = \frac{1}{1 + 1.1 \times 10^{-3}} = 0.9989$$

# find the base transport factor

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$$\alpha_T = \frac{1}{1 + \frac{1}{2} \left( \frac{W_B}{L_{nB}} \right)^2}$$

$$\alpha_T = \frac{1}{1 + 2.0 \times 10^{-4}} = 0.9998$$

# find beta

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$$\beta_F = \frac{\gamma_F \alpha_T}{1 - \gamma_F \alpha_T} = \frac{\alpha_F}{1 - \alpha_F}$$

$$\beta_F = \frac{0.9989 \times 0.9998}{1 - 0.9989 \times 0.9998} = \frac{0.9987}{1 - 0.9987} = 768$$

# find the base current

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$$I_C = 50 \mu\text{A}$$

$$\beta_F = 768$$

$$I_B = I_C / \beta_F$$

$$I_B = I_C / \beta_F = 65 \text{ nA}$$

$$\alpha_F = \frac{\beta_F}{\beta_F + 1} = 0.9987$$



# Forward saturation current density

$$I_C(V_{BE}, V_{BC}) = \alpha_F I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right) - I_{R0} \left( e^{qV_{BC}/k_B T} - 1 \right)$$

$$I_E(V_{BE}, V_{BC}) = I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right) - \alpha_R I_{R0} \left( e^{qV_{BC}/k_B T} - 1 \right)$$

$$I_{F0} = qA \left( \frac{D_{nB}}{W_B} \frac{n_i^2}{N_{AB}} + \frac{D_{pE}}{W_E} \frac{n_i^2}{N_{DE}} \right) \rightarrow qA \left( \frac{D_{nB}}{W_B} \frac{n_i^2}{N_{AB}} + \frac{D_{pE}}{L_{pE}} \frac{n_i^2}{N_{DE}} \right)$$

$$I_{F0} = 1.33 \times 10^{-16} \text{ A}$$

# Reverse saturation current density

$$I_C(V_{BE}, V_{BC}) = \alpha_F I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right) - I_{R0} \left( e^{qV_{BC}/k_B T} - 1 \right)$$

$$I_E(V_{BE}, V_{BC}) = I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right) - \alpha_R I_{R0} \left( e^{qV_{BC}/k_B T} - 1 \right)$$

$$I_{R0} = qA \left( \frac{D_{nB}}{W_B} \frac{n_i^2}{N_{AB}} + \frac{D_{pC}}{W_C} \frac{n_i^2}{N_{DC}} \right)$$

$$I_{R0} = 1.91 \times 10^{-16} \text{ A}$$

# Ebers-Moll parameters summary

$$I_C(V_{BE}, V_{BC}) = \alpha_F I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right) - I_{R0} \left( e^{qV_{BC}/k_B T} - 1 \right)$$

$$I_E(V_{BE}, V_{BC}) = I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right) - \alpha_R I_{R0} \left( e^{qV_{BC}/k_B T} - 1 \right)$$

$$\alpha_R = \alpha_F \frac{I_{F0}}{I_{R0}}$$

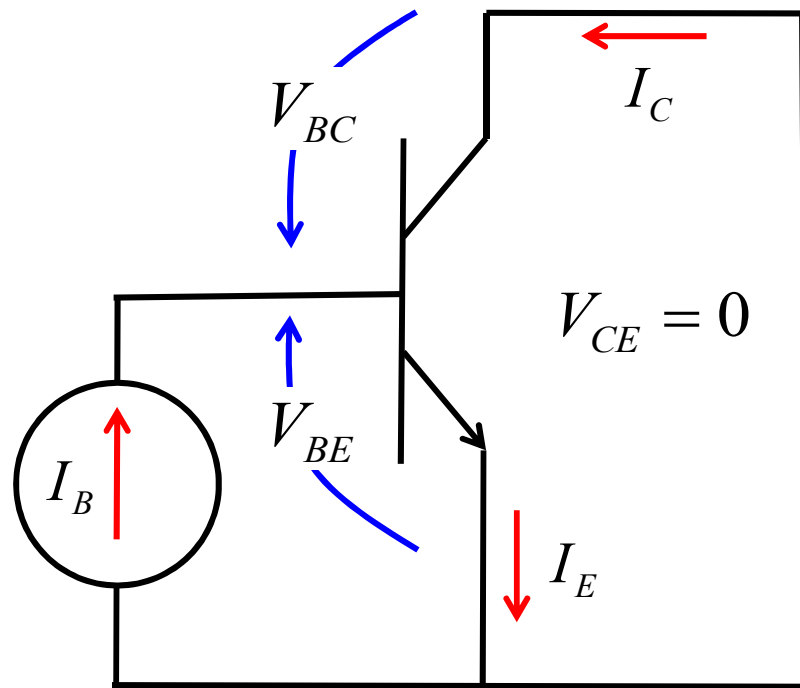
$$\alpha_F = 0.9987$$

$$I_{F0} = 1.33 \times 10^{-16} \text{ A}$$

$$\alpha_R = 0.70 \quad \beta_R = \frac{\alpha_R}{1 - \alpha_R} = 2.3$$

$$I_{R0} = 1.91 \times 10^{-16} \text{ A}$$

# What is $I_C$ ?



$$I_B = 50 \text{ nA}$$

$$I_C = \alpha_F I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right) - I_{R0} \left( e^{qV_{BC}/k_B T} - 1 \right)$$

$$I_E = I_{F0} \left( e^{qV_{BE}/k_B T} - 1 \right) - \alpha_R I_{R0} \left( e^{qV_{BC}/k_B T} - 1 \right)$$

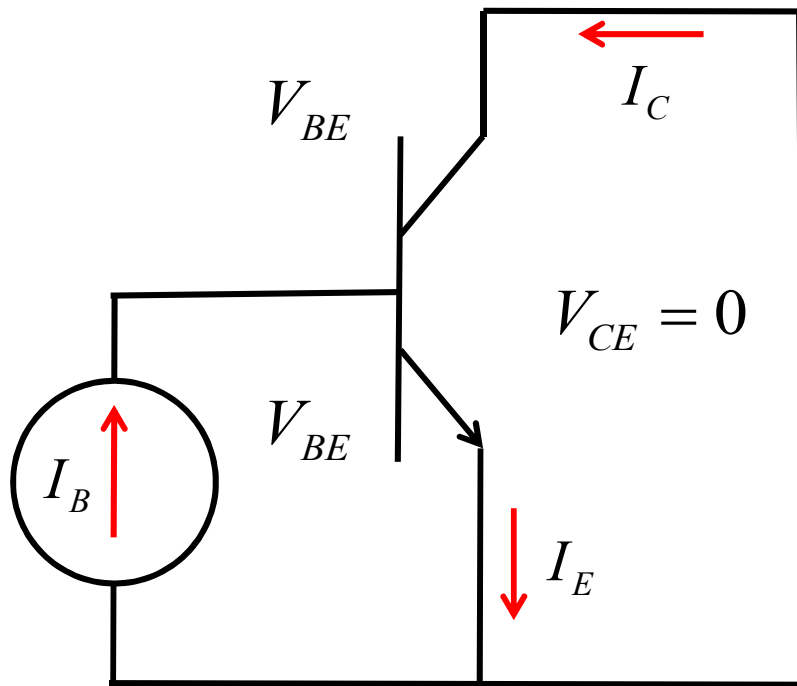
$$V_{CE} = V_{BE} - V_{BC} = 0$$

$$V_{BC} = V_{BE}$$

$$I_C = \left( \alpha_F I_{F0} - I_{R0} \right) \left( e^{qV_{BE}/k_B T} - 1 \right)$$

$$I_E = \left( I_{F0} - \alpha_R I_{R0} \right) \left( e^{qV_{BE}/k_B T} - 1 \right)$$

# What is $I_C$ ?



$$I_B = 50 \text{ nA}$$

$$I_C = (\alpha_F I_{F0} - I_{R0}) (e^{qV_{BE}/k_B T} - 1)$$

$$I_E = (I_{F0} - \alpha_R I_{R0}) (e^{qV_{BE}/k_B T} - 1)$$

$$I_B = I_E - I_C$$

$$I_B = [(1 - \alpha_F) I_{F0} + (1 - \alpha_F) I_{R0}] (e^{qV_{BE}/k_B T} - 1)$$

$$(e^{qV_{BE}/k_B T} - 1) = \frac{I_B}{[(1 - \alpha_F) I_{F0} + (1 - \alpha_F) I_{R0}]}$$

# What is $I_C$ ?

$$I_C = \left\{ \frac{(\alpha_F I_{F0} - I_{R0})}{(1 - \alpha_F) I_{F0} + (1 - \alpha_F) I_{R0}} \right\} I_B$$

$$I_B = 50 \text{ nA}$$

$$I_{F0} = 1.33 \times 10^{-16} \text{ A}$$

$$I_C = -1.01 \times I_B$$

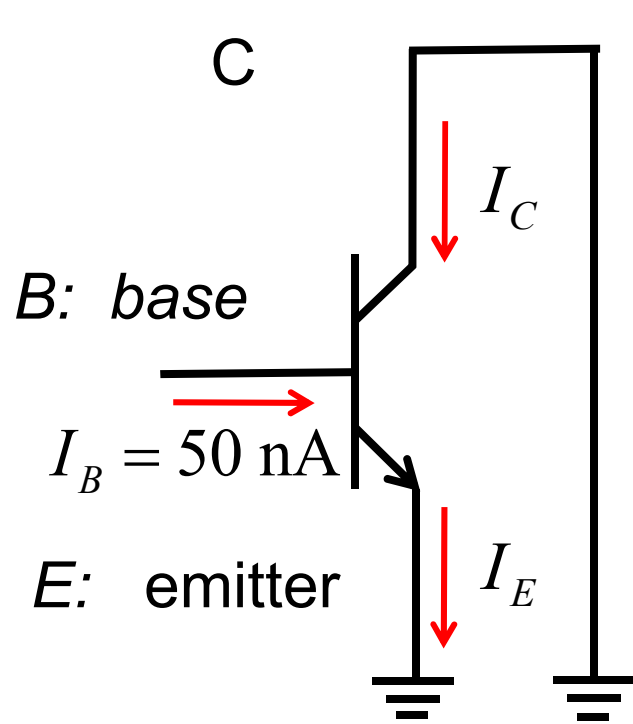
$$I_{R0} = 1.91 \times 10^{-16} \text{ A}$$

$$I_B = -50 \text{ nA}$$

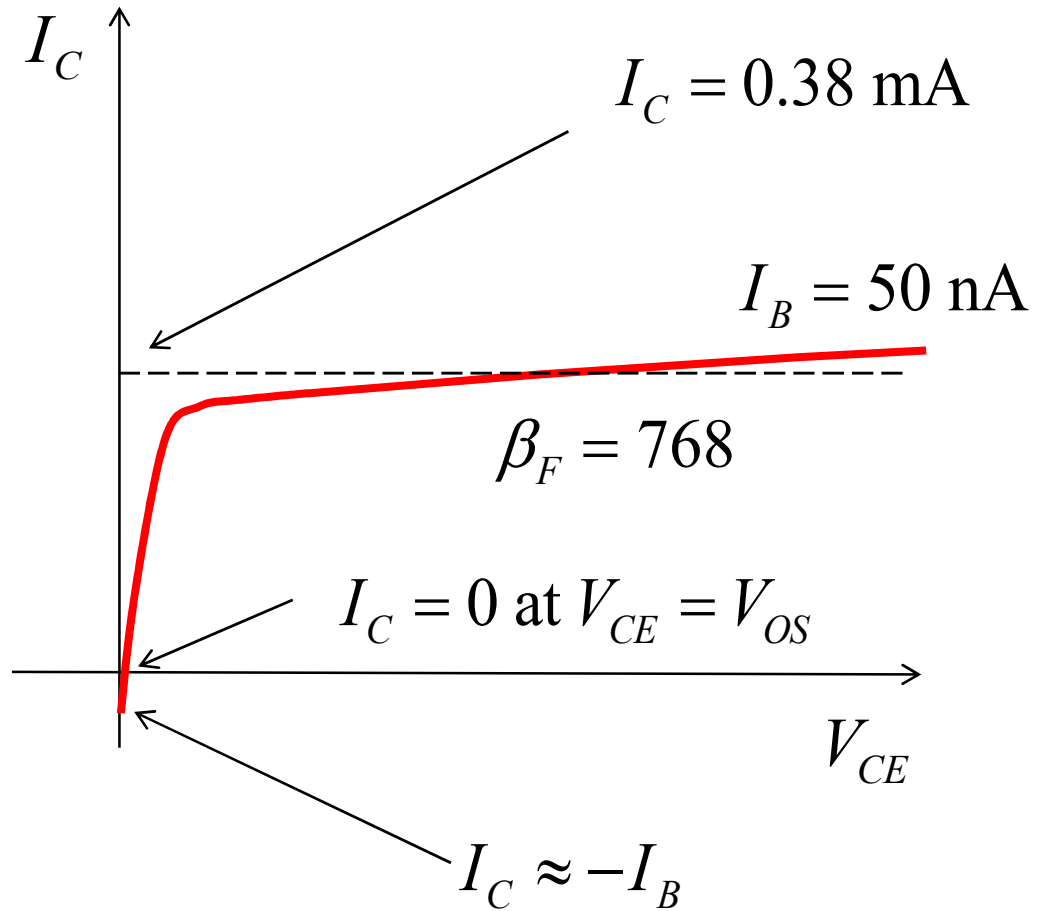
$$\alpha_F = 0.9987$$

$$\alpha_R = 0.70$$

# result



NPN BJT



# Conclusions

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- BJT current is described through a pair of coupled diodes and current sources facing back-to-back: the Ebers-Moll model
- The Ebers-Moll model can be solved for key performance parameters quickly and self-consistently
- Current gain can be very substantial, under the correct circumstances